

EE 330

Lecture 23

- Small Signal Analysis
- Small Signal Analysis of BJT Amplifier

Small-Signal Principle

Goal: to predict circuit performance in the vicinity of an operating point (Q-point)

Will be extended to functions of two and three variables (e.g. BJTs and MOSFETs)

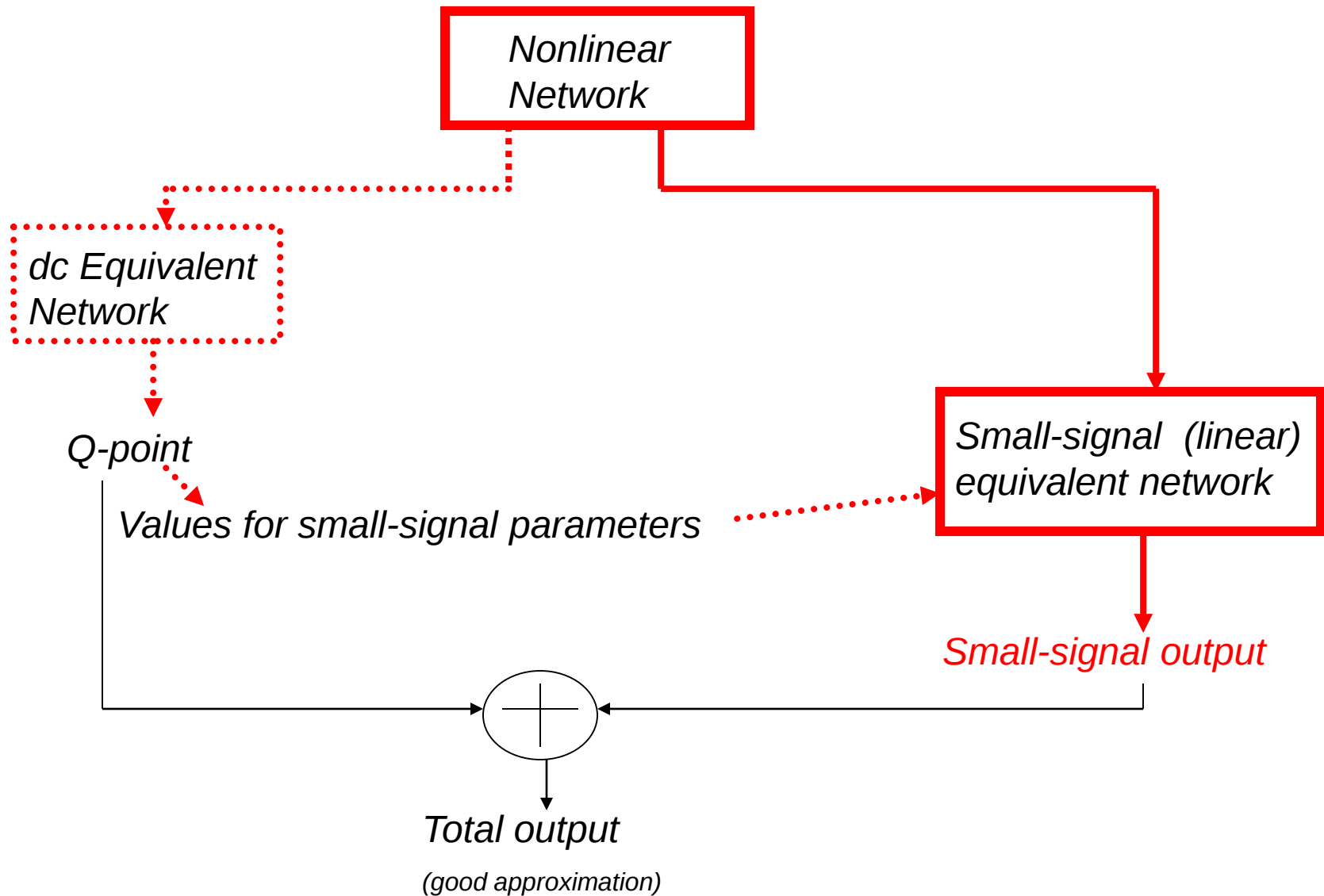
Natural approach to small-signal analysis of nonlinear networks

1. *Solve nonlinear network*
2. *Linearize solution*

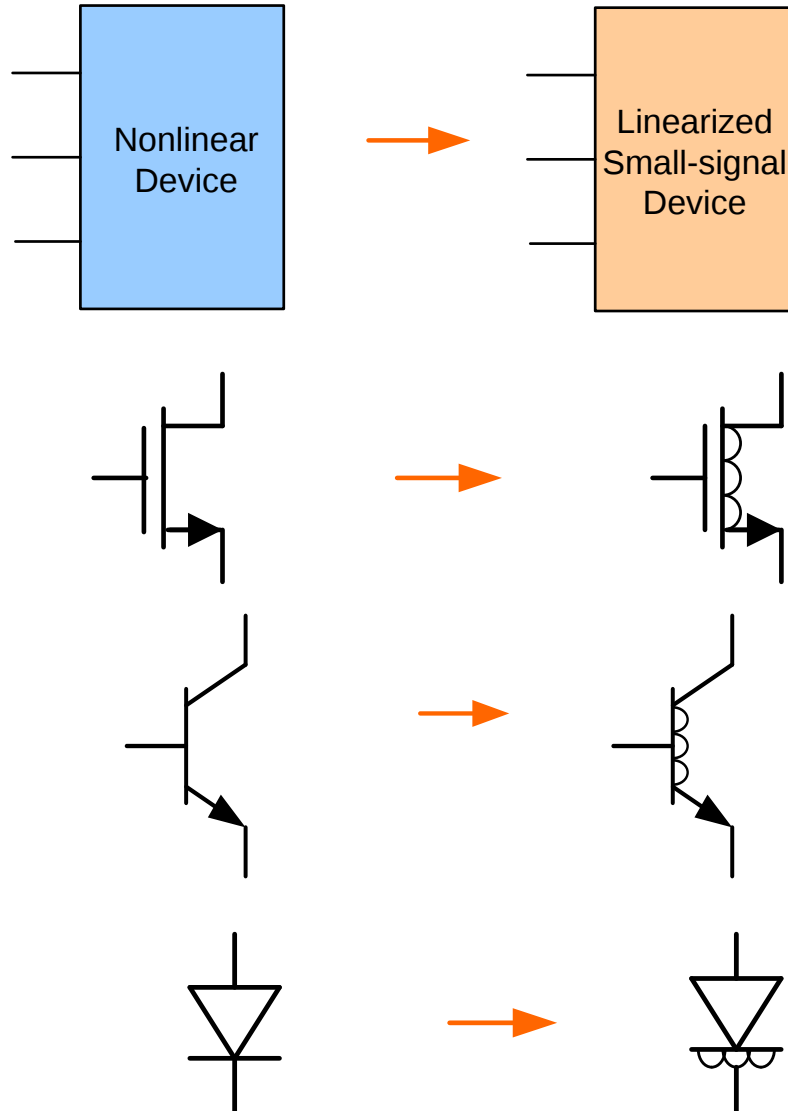
Alternative Approach to small-signal analysis of nonlinear networks

1. *Linearize nonlinear devices (all)*
2. *Obtain small-signal model from linearized device models*
3. *Replace all devices with small-signal equivalent*
4. *Solve linear small-signal network*

“Alternative” Approach to small-signal analysis of nonlinear networks



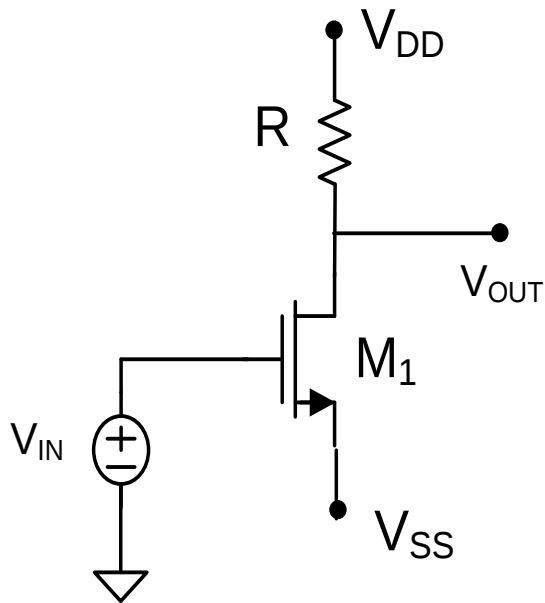
Linearized nonlinear devices



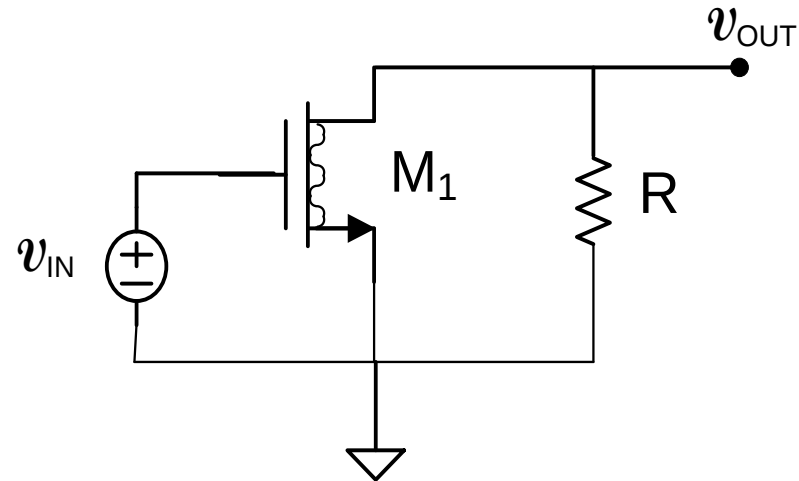
This terminology will be used in THIS course to emphasize difference between nonlinear model and linearized small signal model

Example:

It will be shown that the nonlinear circuit has the linearized small-signal network given



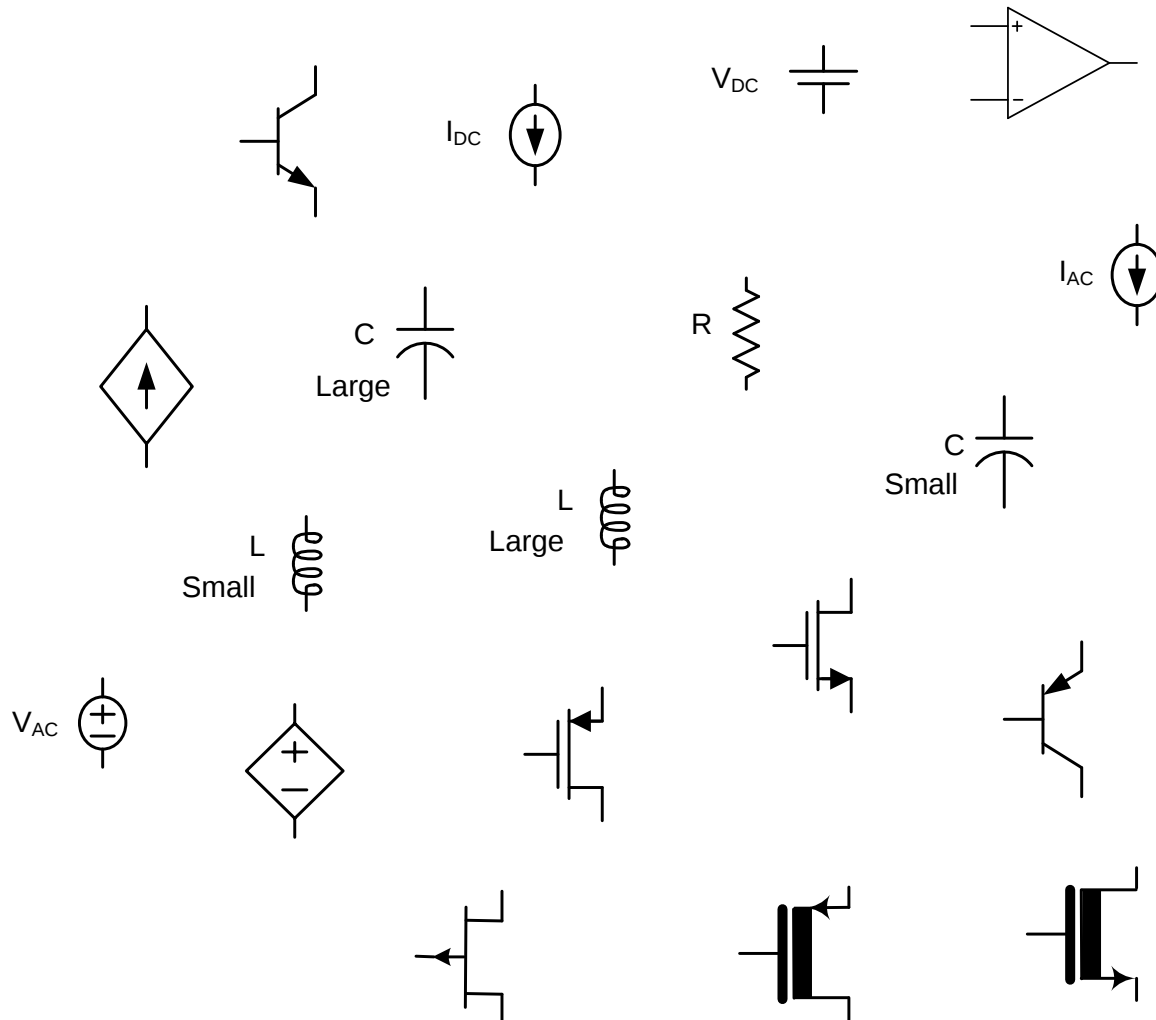
Nonlinear network



Linearized small-signal network

Linearized Small-Signal Circuit Elements

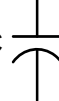
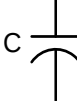
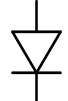
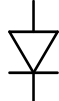
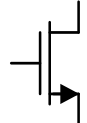
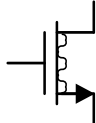
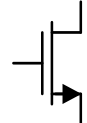
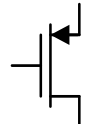
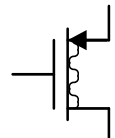
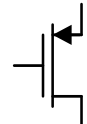
Must obtain the linearized small-signal circuit element for ALL linear and nonlinear circuit elements



(Will also give models that are usually used for Q-point calculations : Simplified dc models)

Small-signal and simplified dc equivalent elements

	Element	ss equivalent	Simplified dc equivalent
dc Voltage Source	V_{DC} 		V_{DC} 
ac Voltage Source	V_{AC} 	V_{AC} 	
dc Current Source	I_{DC} 		I_{DC} 
ac Current Source	I_{AC} 	I_{AC} 	
Resistor	R 	R 	R 

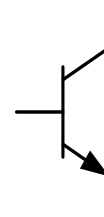
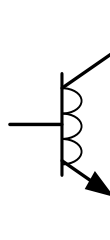
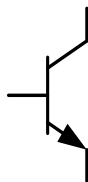
	Element	ss equivalent	Simplified dc equivalent
Capacitors	C Large 		
	C Small 		
Inductors	L Large 		
	L Small 		
Diodes			
MOS transistors (MOSFET (enhancement or depletion), JFET)			
			

Element

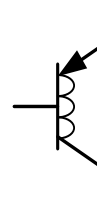
ss equivalent

Simplified dc
equivalent

Bipolar
Transistors



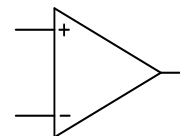
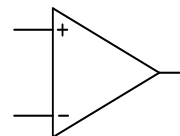
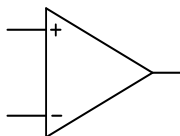
Simplified



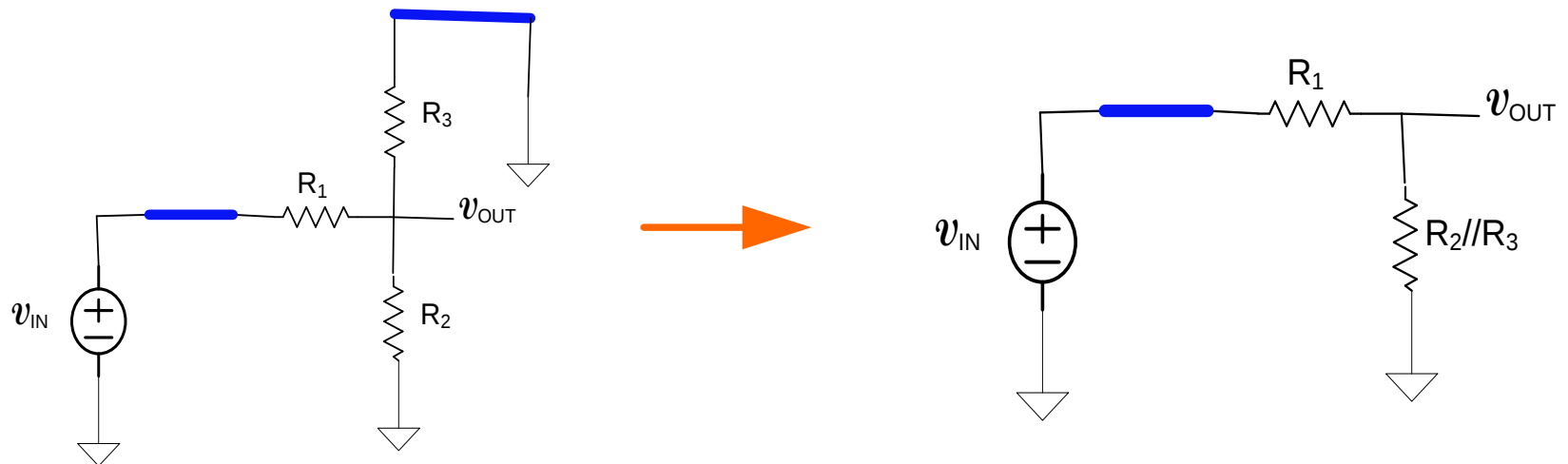
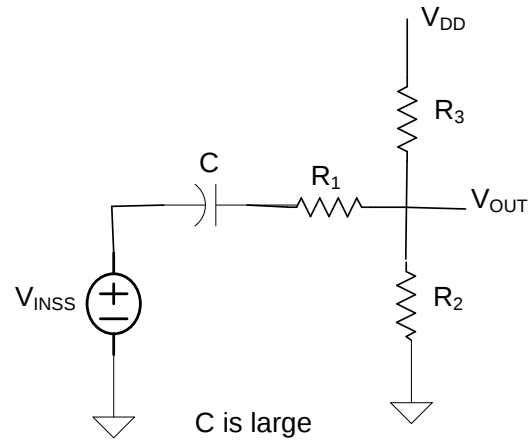
Simplified

Dependent
Sources
(Linear)

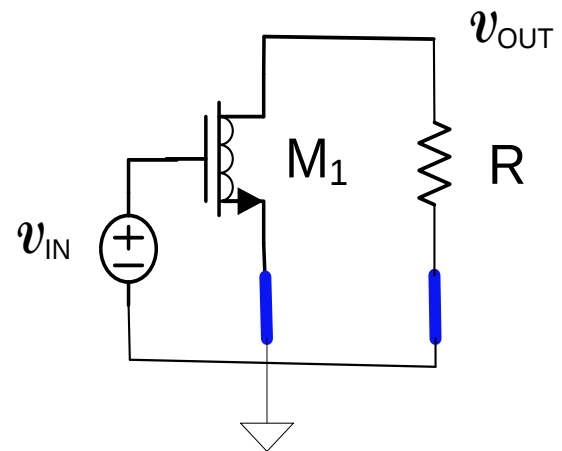
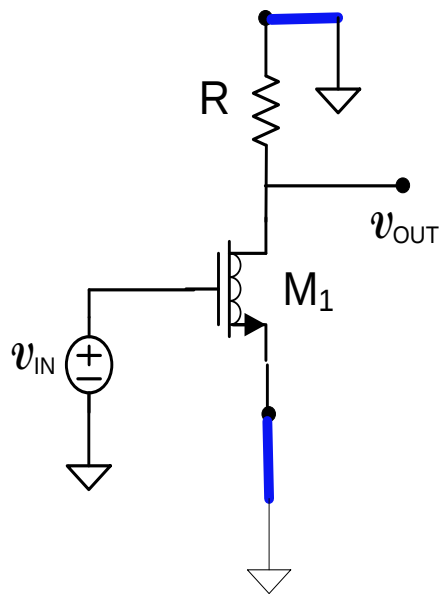
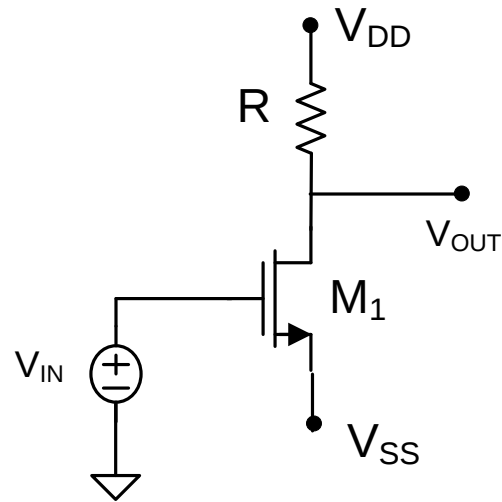
$$\begin{array}{ll} V_O = A_V V_{IN} & I_O = A_I I_{IN} \\ V_O = R_T I_{IN} & I_O = G_T V_{IN} \end{array}$$



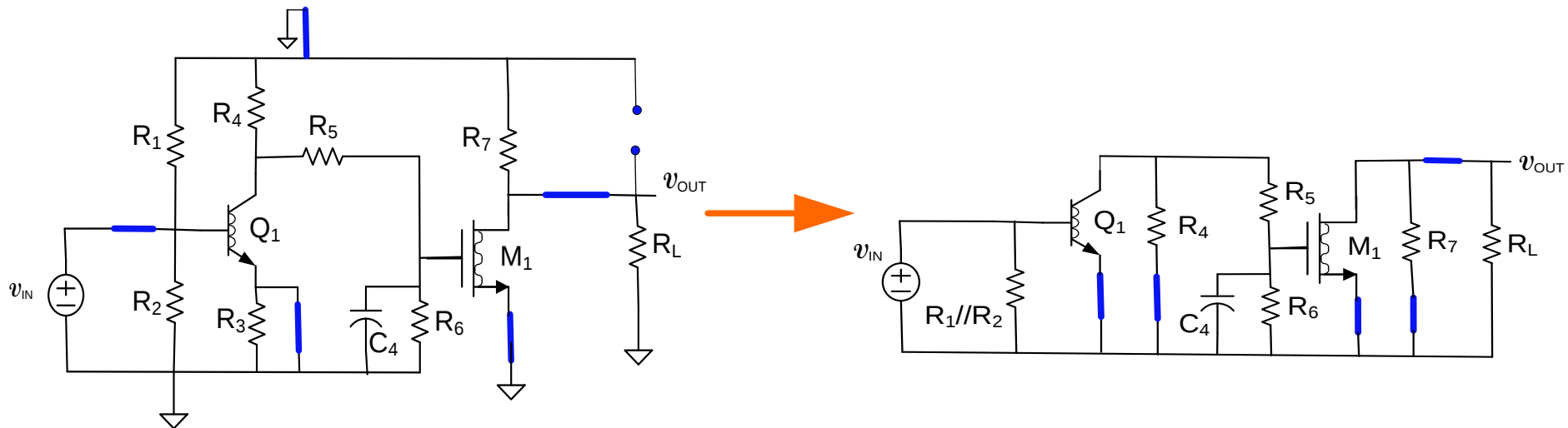
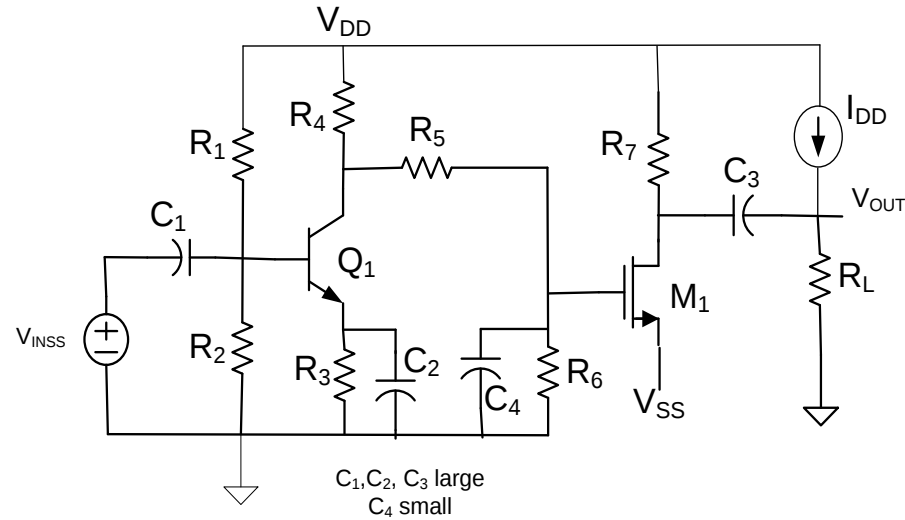
Example: Obtain the small-signal equivalent circuit



Example: Obtain the small-signal equivalent circuit

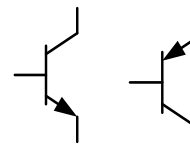
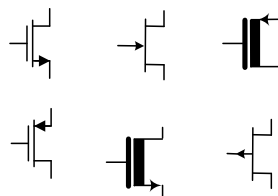
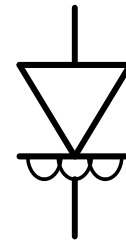
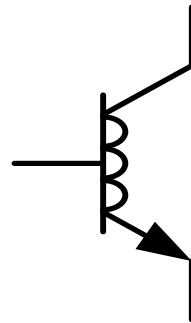
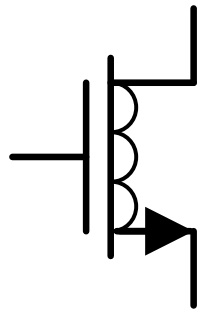


Example: Obtain the small-signal equivalent circuit

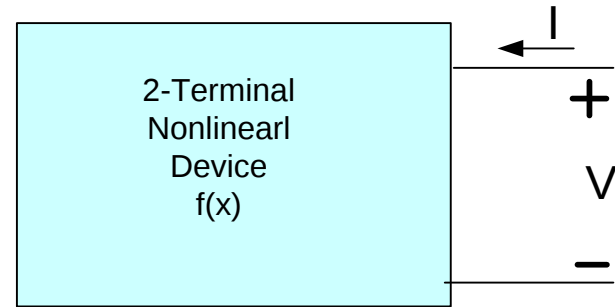


How is the small-signal equivalent circuit obtained from the nonlinear circuit?

What is the small-signal equivalent of the MOSFET, BJT, and diode ?



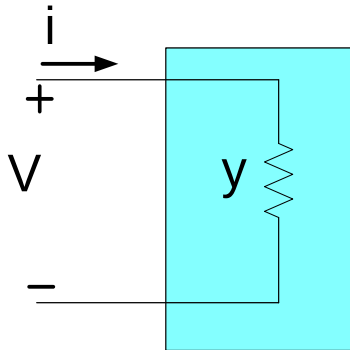
Small-Signal Diode Model



$$y = \left. \frac{\partial I}{\partial V} \right|_{V=V_Q}$$

$$i = y v$$

A Small Signal Equivalent Circuit



Thus, for the diode

$$R_d = \left(\frac{\partial I_D}{\partial V_D} \right)^{-1}_Q$$

Small-Signal Diode Model

For the diode

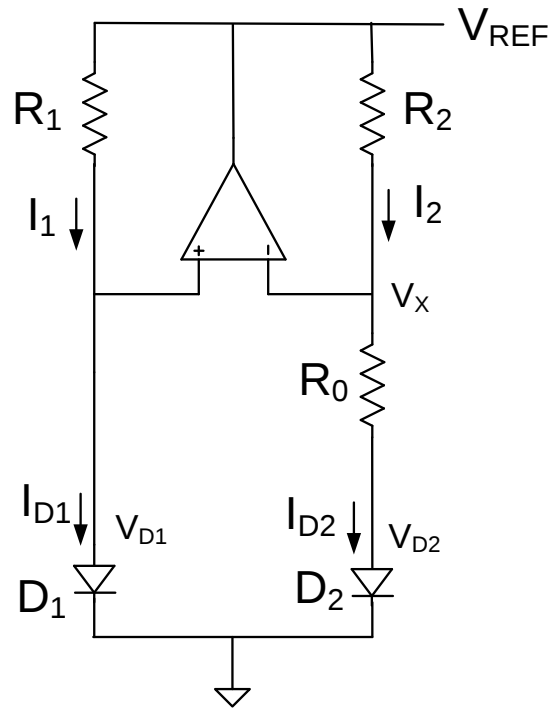


$$I_D = I_S e^{V_D / V_t}$$

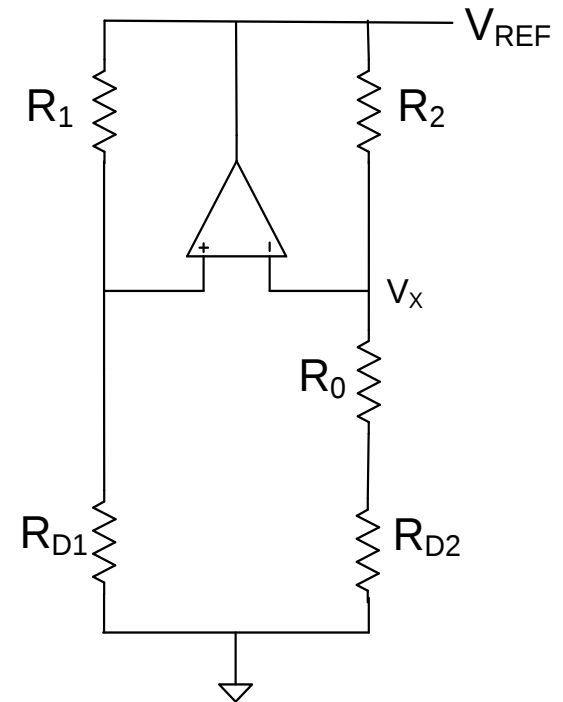
$$G_d = \left. \frac{\partial I_D}{\partial V_D} \right|_Q = \left[\left(I_S e^{V_D / V_t} \right) \frac{1}{V_t} \right]_Q = \frac{I_{DQ}}{V_t}$$

$$R_d = \frac{V_t}{I_{DQ}}$$

Example of diode circuit where small-signal diode model is useful



Voltage Reference

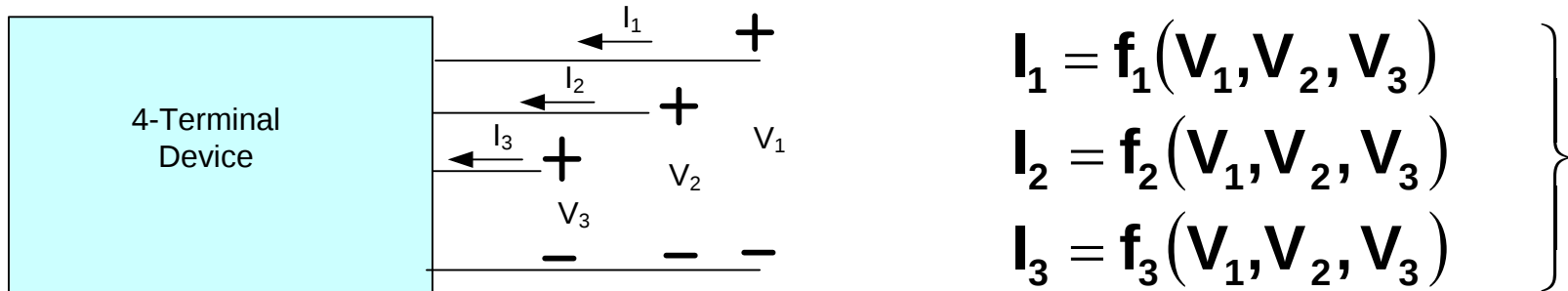


Small-signal model of
Voltage Reference
(useful for compensation
when parasitic Cs included)

Small-Signal Model of BJT and MOSFET

Consider 4-terminal network

(3-terminal and 2-terminal and 1-terminal devices then become special cases)



4 different ways to choose reference terminal

Six port electrical variables $\{I_1, I_2, I_3, V_1, V_2, V_3\}$

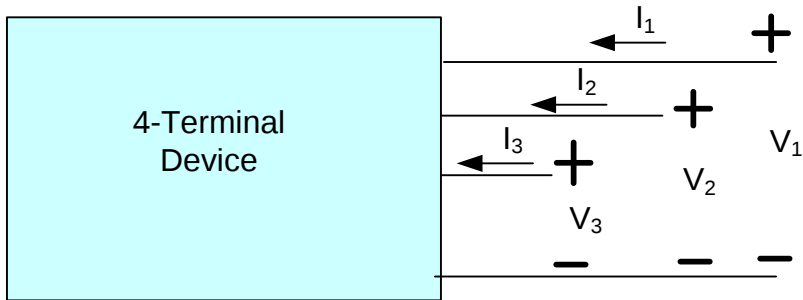
Number of ways to choose independent variables $\binom{6}{3} = \frac{6!}{(6-3)!3!} = 20$

Number of potentially different ways to represent same device 80

We will choose one of these 80 which uses port voltages as independent variables

Small-Signal Model of BJT and MOSFET

Consider 4-terminal network



$$\left. \begin{aligned} I_1 &= f_1(V_1, V_2, V_3) \\ I_2 &= f_2(V_1, V_2, V_3) \\ I_3 &= f_3(V_1, V_2, V_3) \end{aligned} \right\}$$

Define

$$i_1 = I_1 - I_{1Q}$$

$$i_2 = I_2 - I_{2Q}$$

$$i_3 = I_3 - I_{3Q}$$

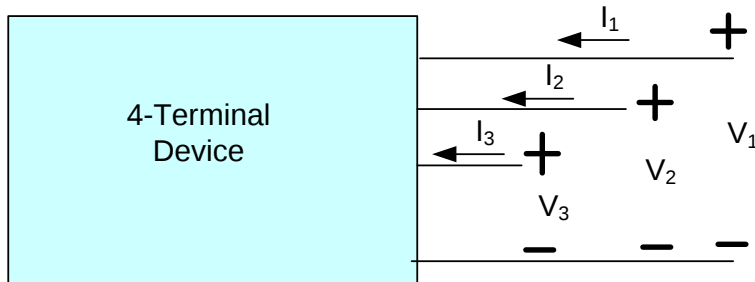
$$u_1 = V_1 - V_{1Q}$$

$$u_2 = V_2 - V_{2Q}$$

$$u_3 = V_3 - V_{3Q}$$

Small signal model is that which represents the relationship between the small signal voltages and the small signal currents

Small-Signal Model of 4-Terminal Network



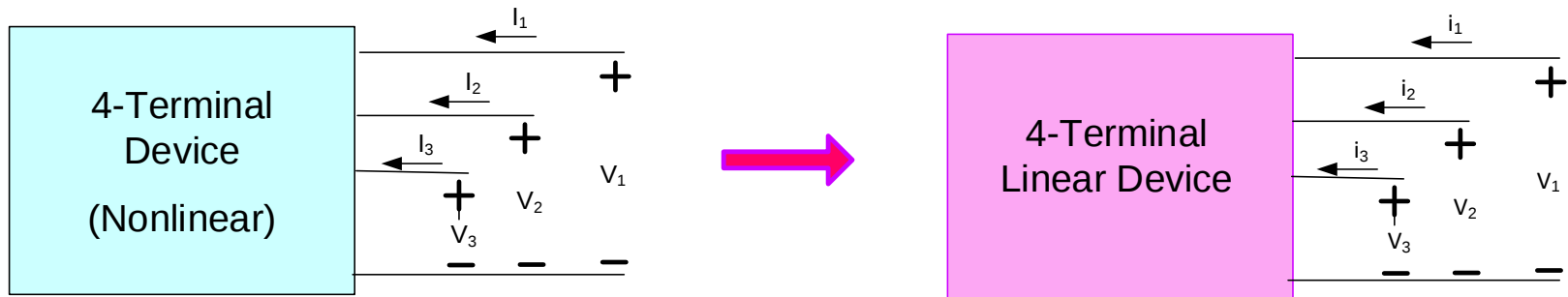
$$\left. \begin{aligned} i_1 &= g_1(v_1, v_2, v_3) \\ i_2 &= g_2(v_1, v_2, v_3) \\ i_3 &= g_3(v_1, v_2, v_3) \end{aligned} \right\}$$

Small signal model is that which represents the relationship between the small signal voltages and the small signal currents

For small signals, this relationship should be linear

Can be thought of as a change in coordinate systems from the large signal coordinate system to the small-signal coordinate system

Small-Signal Model of 4-Terminal Network

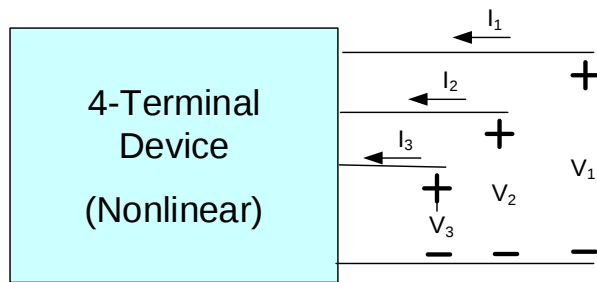


$$\left. \begin{aligned} I_1 &= f_1(V_1, V_2, V_3) \\ I_2 &= f_2(V_1, V_2, V_3) \\ I_3 &= f_3(V_1, V_2, V_3) \end{aligned} \right\}$$

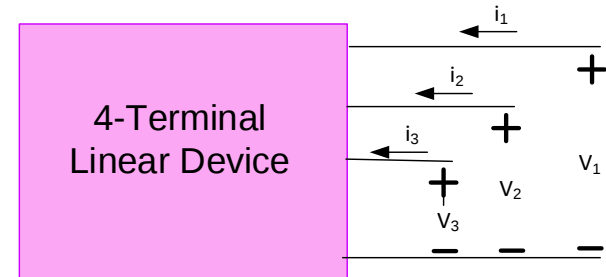
$$\left. \begin{aligned} i_1 &= g_1(v_1, v_2, v_3) \\ i_2 &= g_2(v_1, v_2, v_3) \\ i_3 &= g_3(v_1, v_2, v_3) \end{aligned} \right\}$$

Mapping is unique (with same models)

Small-Signal Model of 4-Terminal Network



$$\left. \begin{aligned} I_1 &= f_1(V_1, V_2, V_3) \\ I_2 &= f_2(V_1, V_2, V_3) \\ I_3 &= f_3(V_1, V_2, V_3) \end{aligned} \right\}$$



$$\left. \begin{aligned} i_1 &= g_1(v_1, v_2, v_3) \\ i_2 &= g_2(v_1, v_2, v_3) \\ i_3 &= g_3(v_1, v_2, v_3) \end{aligned} \right\}$$

Does inverse mapping exist?

Yes

Is it unique (with same models)?

No

Multiple nonlinear circuits can have same small-signal circuit

Recall for a function of one variable

$$y = f(x)$$

Taylor's Series Expansion about the point x_0 (x_0 is termed the expansion point or the Q-point)

$$y = f(x) = f(x)|_{x=x_0} + \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x - x_0) + \left. \frac{\partial^2 f}{\partial x^2} \right|_{x=x_0} \frac{1}{2!} (x - x_0)^2 + \dots$$

If $x - x_0$ is small

$$y \cong f(x)|_{x=x_0} + \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x - x_0)$$

$$y \cong y_0 + \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x - x_0)$$

Recall for a function of one variable

$$y = f(x)$$

If $x - x_0$ is small

$$y \cong y_0 + \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x - x_0)$$

$$y - y_0 = \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x - x_0)$$

If we define the small signal variables as

$$\boldsymbol{y} = y - y_0$$

$$\boldsymbol{x} = x - x_0$$

Recall for a function of one variable

$$y = f(x)$$

If $x - x_0$ is small

$$y \cong y_0 + \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x - x_0)$$

$$y - y_0 = \left. \frac{\partial f}{\partial x} \right|_{x=x_0} (x - x_0)$$

If we define the small signal variables as

$$\boldsymbol{y} = y - y_0$$

$$\boldsymbol{x} = x - x_0$$

Then

$$\boldsymbol{y} = \left. \frac{\partial f}{\partial x} \right|_{x=x_0} \bullet \boldsymbol{x}$$

This relationship is linear !

Consider now a function of n variables

$$y = f(x_1, \dots, x_n) = f(\bar{x})$$

If we consider an arbitrary expansion point $\bar{x}_0 = \{x_{10}, x_{20}, \dots, x_{n0}\}$

The multivariate Taylor's series expansion around the point $\bar{x}_0$ is given by

$$y = f(\bar{x}) = f(\mathbf{x}) \Big|_{\bar{x}=\bar{x}_0} + \sum_{k=1}^n \left(\frac{\partial f}{\partial x_k} \Big|_{\bar{x}=\bar{x}_0} (x_k - x_{k0}) \right) \\ + \sum_{\substack{k=1 \\ j=1}}^{n,n} \frac{\partial^2 f}{\partial x_j \partial x_k} \Big|_{\bar{x}=\bar{x}_0} \frac{1}{2!} (x_j - x_{j0})(x_k - x_{k0}) + \dots (\text{H.O.T.})$$

Truncating after first-order terms, we obtain the approximation

$$y - y_0 \cong \sum_{k=1}^n \left(\frac{\partial f}{\partial x_k} \Big|_{\bar{x}=\bar{x}_0} (x_k - x_{k0}) \right)$$

where $y_0 = f(\mathbf{x}) \Big|_{\bar{x}=\bar{x}_0}$

Multivariate Taylors Series Expansion

$$y = f(x_1, \dots, x_n) = f(\bar{x})$$

Linearized approximation

$$y - y_0 \cong \sum_{k=1}^n \left(\frac{\partial f}{\partial x_k} \bigg|_{\bar{x} = \bar{x}_0} (x_k - x_{k0}) \right)$$

This can be expressed as

$$y_{ss} \cong \sum_{k=1}^n a_k x_{ss_k} \qquad \left(y \cong \sum_{k=1}^n a_k x_k \right)$$

where

$$y_{ss} = y - y_0 \qquad (y = y - y_0)$$

$$x_{ss_k} = x_k - x_{k0} \qquad (x_k = x_k - x_{k0})$$

$$a_k = \frac{\partial f}{\partial x_k} \bigg|_{\bar{x} = \bar{x}_0}$$

In the more general form¹, the multivariate Taylor's series expansion can be expressed as

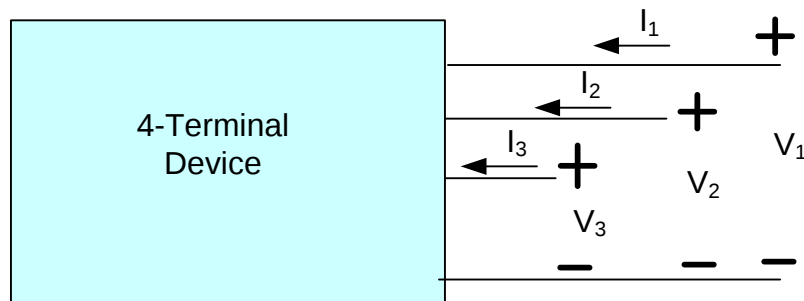
$$f(x_1, \dots, x_n) = \alpha_o + \sum_{m=1}^{\infty} \left(\sum_{\substack{k_1, \dots, k_n \\ \left(\sum_j k_j = m \right)}} \alpha_{k_1, \dots, k_n; m} (x_1 - x_{1,o})^{k_1} \dots (x_n - x_{n,o})^{k_n} \right) \quad (7)$$

$$\alpha_o = f(x_{1o}, \dots, x_{no})$$

$$\alpha_{k_1, \dots, k_n; m} = \frac{1}{k_1! \dots k_n!} \frac{\partial^m f}{\partial^{k_1} x_1 \dots \partial^{k_n} x_n} \bigg|_{x_{1o}, \dots, x_{no}} \quad (8)$$

¹ <http://www.chem.mtu.edu/~tbco/cm416/taylor.html>

Consider 4-terminal network



$$\left. \begin{aligned} I_1 &= f_1(V_1, V_2, V_3) \\ I_2 &= f_2(V_1, V_2, V_3) \\ I_3 &= f_3(V_1, V_2, V_3) \end{aligned} \right\}$$

Nonlinear network characterized by 3 functions each functions of 3 variables

Consider now 3 functions each functions of 3 variables

$$\left. \begin{aligned} \mathbf{l}_1 &= \mathbf{f}_1(\mathbf{v}_1, \mathbf{v}_2, \mathbf{v}_3) \\ \mathbf{l}_2 &= \mathbf{f}_2(\mathbf{v}_1, \mathbf{v}_2, \mathbf{v}_3) \\ \mathbf{l}_3 &= \mathbf{f}_3(\mathbf{v}_1, \mathbf{v}_2, \mathbf{v}_3) \end{aligned} \right\}$$

Define

$$\vec{\mathbf{V}}_Q = \begin{bmatrix} \mathbf{V}_{1Q} \\ \mathbf{V}_{2Q} \\ \mathbf{V}_{3Q} \end{bmatrix}$$

In what follows, we will use $\vec{\mathbf{V}}_Q$ as an expansion point in a Taylor's series expansion.

Consider now 3 functions each functions of 3 variables

$$\left. \begin{aligned} \mathbf{I}_1 &= \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3) \\ \mathbf{I}_2 &= \mathbf{f}_2(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3) \\ \mathbf{I}_3 &= \mathbf{f}_3(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3) \end{aligned} \right\} \quad \text{Define} \quad \bar{\mathbf{V}}_Q = \begin{bmatrix} \mathbf{V}_{1Q} \\ \mathbf{V}_{2Q} \\ \mathbf{V}_{3Q} \end{bmatrix}$$

Consider first the function $\mathbf{I}_1$

The multivariate Taylors Series expansion of $\mathbf{I}_1$, around the operating point $\bar{\mathbf{V}}_Q$. when truncated after first-order terms, can be expressed as:

$$\mathbf{I}_1 = \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3) \cong \mathbf{f}_1(\mathbf{V}_{1Q}, \mathbf{V}_{2Q}, \mathbf{V}_{3Q}) + \left. \frac{\partial \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_1} \right|_{\bar{\mathbf{V}}=\bar{\mathbf{V}}_Q} (\mathbf{V}_1 - \mathbf{V}_{1Q}) + \left. \frac{\partial \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_2} \right|_{\bar{\mathbf{V}}=\bar{\mathbf{V}}_Q} (\mathbf{V}_2 - \mathbf{V}_{2Q}) + \left. \frac{\partial \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_3} \right|_{\bar{\mathbf{V}}=\bar{\mathbf{V}}_Q} (\mathbf{V}_3 - \mathbf{V}_{3Q})$$

or equivalently as:

$$\mathbf{I}_1 - \mathbf{I}_{1Q} = \left. \frac{\partial \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_1} \right|_{\bar{\mathbf{V}}=\bar{\mathbf{V}}_Q} (\mathbf{V}_1 - \mathbf{V}_{1Q}) + \left. \frac{\partial \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_2} \right|_{\bar{\mathbf{V}}=\bar{\mathbf{V}}_Q} (\mathbf{V}_2 - \mathbf{V}_{2Q}) + \left. \frac{\partial \mathbf{f}_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_3} \right|_{\bar{\mathbf{V}}=\bar{\mathbf{V}}_Q} (\mathbf{V}_3 - \mathbf{V}_{3Q})$$

repeating from previous slide:

$$I_1 - I_{1Q} = \left. \frac{\partial f_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_1} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q} (\mathbf{V}_1 - \mathbf{V}_{1Q}) + \left. \frac{\partial f_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_2} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q} (\mathbf{V}_2 - \mathbf{V}_{2Q}) + \left. \frac{\partial f_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_3} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q} (\mathbf{V}_3 - \mathbf{V}_{3Q})$$

Make the following definitions

$$y_{11} = \left. \frac{\partial f_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_1} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

$$y_{12} = \left. \frac{\partial f_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_2} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

$$y_{13} = \left. \frac{\partial f_1(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_3} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

$$\tilde{i}_1 = I_1 - I_{1Q}$$

$$\tilde{i}_2 = I_2 - I_{2Q}$$

$$\tilde{i}_3 = I_3 - I_{3Q}$$

$$u_1 = \mathbf{V}_1 - \mathbf{V}_{1Q}$$

$$u_2 = \mathbf{V}_2 - \mathbf{V}_{2Q}$$

$$u_3 = \mathbf{V}_3 - \mathbf{V}_{3Q}$$

It thus follows that

$$\tilde{i}_1 = y_{11} u_1 + y_{12} u_2 + y_{13} u_3$$

This is a linear relationship between the small signal electrical variables !

Small Signal Model Development

Nonlinear Model

Linear Model at $\bar{\mathbf{V}}_Q$
(alt. small signal model)

$$\begin{aligned} I_1 &= f_1(V_1, V_2, V_3) \\ I_2 &= f_2(V_1, V_2, V_3) \\ I_3 &= f_3(V_1, V_2, V_3) \end{aligned} \quad \rightarrow \quad \begin{aligned} i_1 &= y_{11}u_1 + y_{12}u_2 + y_{13}u_3 \end{aligned}$$

Extending this approach to the two nonlinear functions I_2 and I_3

$$i_2 = y_{21}u_1 + y_{22}u_2 + y_{23}u_3$$

$$i_3 = y_{31}u_1 + y_{32}u_2 + y_{33}u_3$$

where

$$y_{ij} = \left. \frac{\partial f_i(V_1, V_2, V_3)}{\partial V_j} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

Small Signal Model Development

Nonlinear Model

Linear Model at $\bar{\mathbf{V}}_Q$
(alt. small signal model)

$$\begin{aligned} I_1 &= f_1(V_1, V_2, V_3) \rightarrow i_1 = y_{11}u_1 + y_{12}u_2 + y_{13}u_3 \\ I_2 &= f_2(V_1, V_2, V_3) \rightarrow i_2 = y_{21}u_1 + y_{22}u_2 + y_{23}u_3 \\ I_3 &= f_3(V_1, V_2, V_3) \rightarrow i_3 = y_{31}u_1 + y_{32}u_2 + y_{33}u_3 \end{aligned}$$

where

$$y_{ij} = \left. \frac{\partial f_i(V_1, V_2, V_3)}{\partial V_j} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

Small Signal Model

$$\dot{i}_1 = y_{11}u_1 + y_{12}u_2 + y_{13}u_3$$

$$\dot{i}_2 = y_{21}u_1 + y_{22}u_2 + y_{23}u_3$$

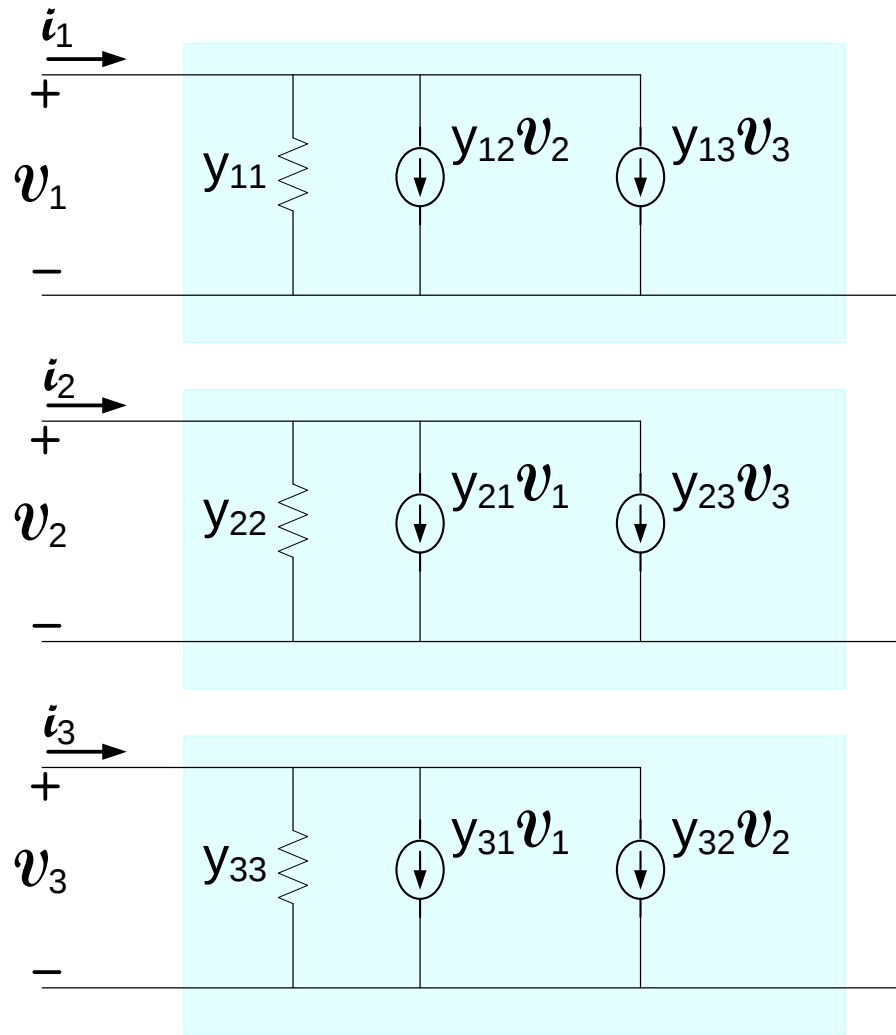
$$\dot{i}_3 = y_{31}u_1 + y_{32}u_2 + y_{33}u_3$$

where

$$y_{ij} = \left. \frac{\partial f_i(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial \mathbf{V}_j} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

- This is a small-signal model of a 4-terminal network and it is linear
- 9 small-signal parameters characterize the linear 4-terminal network
- Small-signal model parameters dependent upon Q-point !
- Termed the y-parameter model or “admittance” –parameter model

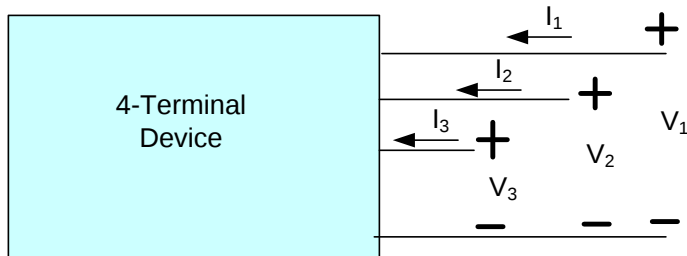
A small-signal equivalent circuit of a 4-terminal nonlinear network (equivalent circuit because has exactly the same port equations)



$$y_{ij} = \left. \frac{\partial f_i(\mathbf{V}_1, \mathbf{V}_2, \mathbf{V}_3)}{\partial V_j} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

Equivalent circuit is not unique
Equivalent circuit is a three-port network

4-terminal small-signal network summary



$$\left. \begin{aligned} I_1 &= f_1(V_1, V_2, V_3) \\ I_2 &= f_2(V_1, V_2, V_3) \\ I_3 &= f_3(V_1, V_2, V_3) \end{aligned} \right\}$$

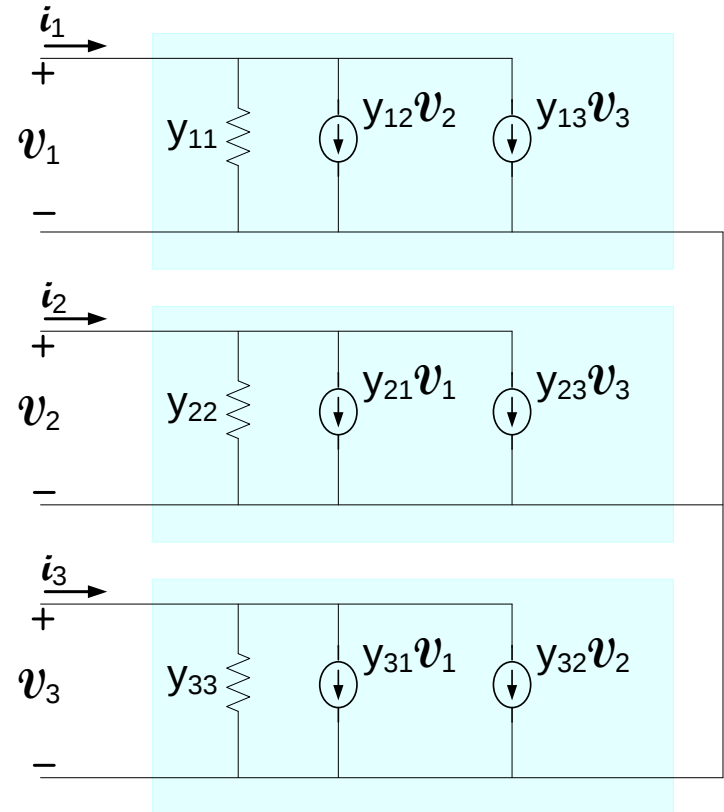
Small signal model:

$$i_1 = y_{11}v_1 + y_{12}v_2 + y_{13}v_3$$

$$i_2 = y_{21}v_1 + y_{22}v_2 + y_{23}v_3$$

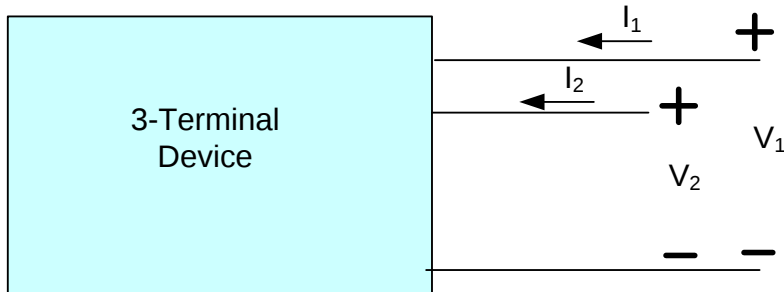
$$i_3 = y_{31}v_1 + y_{32}v_2 + y_{33}v_3$$

$$y_{ij} = \left. \frac{\partial f_i(V_1, V_2, V_3)}{\partial V_j} \right|_{\bar{V} = \bar{V}_Q}$$



Consider 3-terminal network

Small-Signal Model



$$\left. \begin{aligned} I_1 &= f_1(V_1, V_2) \\ I_2 &= f_2(V_1, V_2) \end{aligned} \right\}$$

Define

$$i_1 = I_1 - I_{1Q}$$

$$v_1 = V_1 - V_{1Q}$$

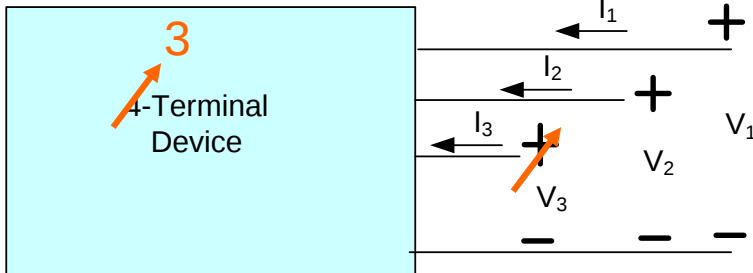
$$i_2 = I_2 - I_{2Q}$$

$$v_2 = V_2 - V_{2Q}$$

Small signal model is that which represents the relationship between the small signal voltages and the small signal currents

Consider 3-terminal network

Small-Signal Model



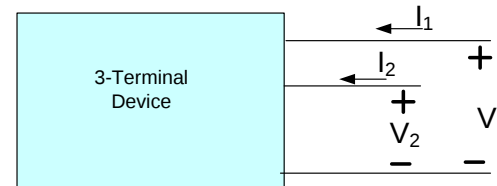
$$\left. \begin{aligned} i_1 &= g_1(v_1, v_2, v_3) \\ i_2 &= g_2(v_1, v_2, v_3) \\ i_3 &= g_3(v_1, v_2, v_3) \end{aligned} \right\}$$

$$\begin{aligned} i_1 &= y_{11}v_1 + y_{12}v_2 + y_{13}v_3 \\ i_2 &= y_{21}v_1 + y_{22}v_2 + y_{23}v_3 \\ i_3 &= y_{31}v_1 + y_{32}v_2 + y_{33}v_3 \end{aligned}$$

$$y_{ij} = \left. \frac{\partial f_i(\mathbf{v}_1, \mathbf{v}_2, \mathbf{v}_3)}{\partial \mathbf{v}_j} \right|_{\bar{\mathbf{v}} = \bar{\mathbf{v}}_Q}$$

Consider 3-terminal network

Small-Signal Model

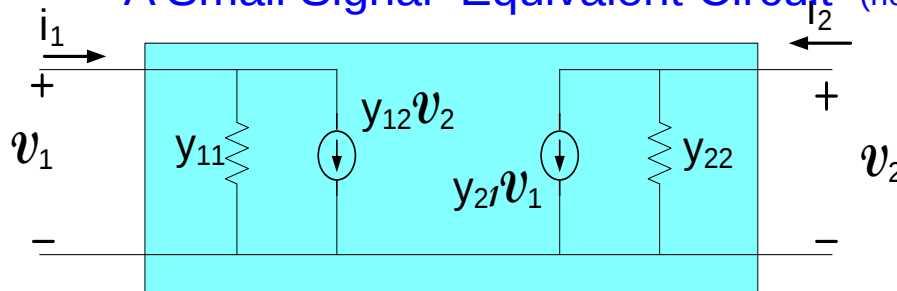


$$\begin{aligned} i_1 &= y_{11}v_1 + y_{12}v_2 \\ i_2 &= y_{21}v_1 + y_{22}v_2 \end{aligned}$$

$$y_{ij} = \left. \frac{\partial f_i(\mathbf{V}_1, \mathbf{V}_2)}{\partial \mathbf{V}_j} \right|_{\bar{\mathbf{V}} = \bar{\mathbf{V}}_Q}$$

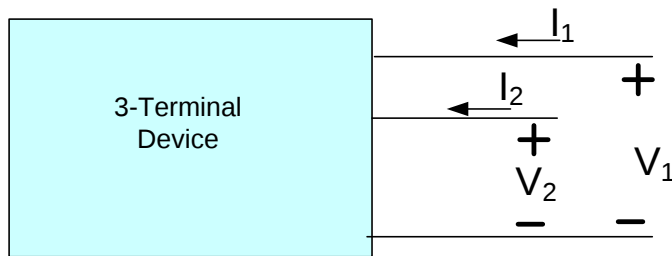
$$\bar{\mathbf{V}} = \begin{pmatrix} V_{1Q} \\ V_{2Q} \end{pmatrix}$$

A Small Signal Equivalent Circuit (not unique)



- Small-signal model is a “two-port”
- 4 small-signal parameters characterize this 3-terminal linear network
- Small signal parameters dependent upon Q-point

3-terminal small-signal network summary

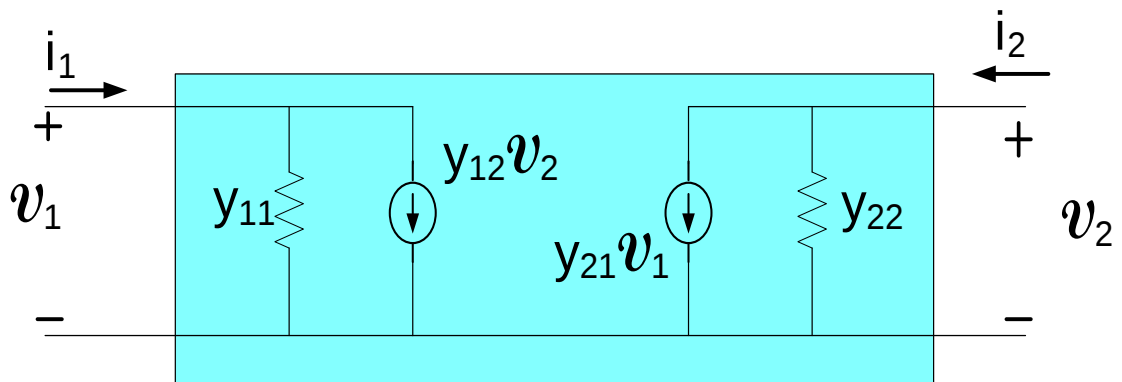


$$\left. \begin{aligned} I_1 &= f_1(V_1, V_2) \\ I_2 &= f_2(V_1, V_2) \end{aligned} \right\}$$

Small signal model:

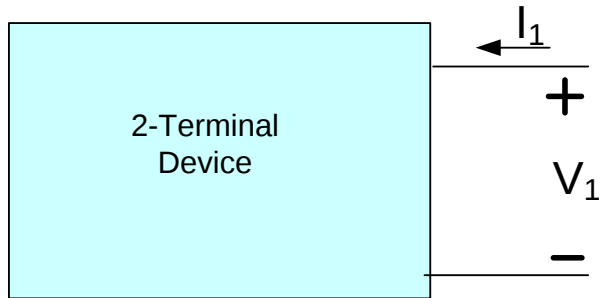
$$\begin{aligned} i_1 &= y_{11}v_1 + y_{12}v_2 \\ i_2 &= y_{21}v_1 + y_{22}v_2 \end{aligned}$$

$$y_{ij} = \left. \frac{\partial f_i(V_1, V_2)}{\partial V_j} \right|_{\bar{V}=\bar{V}_Q}$$



Consider 2-terminal network

Small-Signal Model



$$I_1 = f_1(V_1)$$

Define

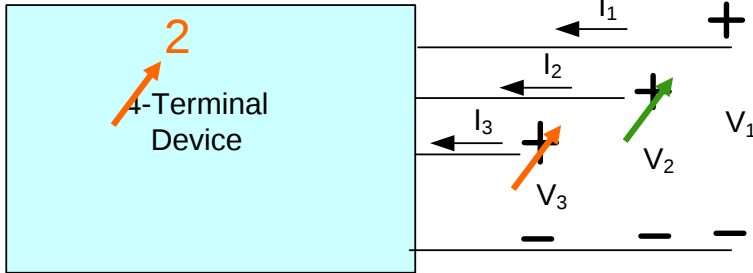
$$i_1 = I_1 - I_{1Q}$$

$$v_1 = V_1 - V_{1Q}$$

Small signal model is that which represents the relationship between the small signal voltages and the small signal currents

Consider 2-terminal network

Small-Signal Model

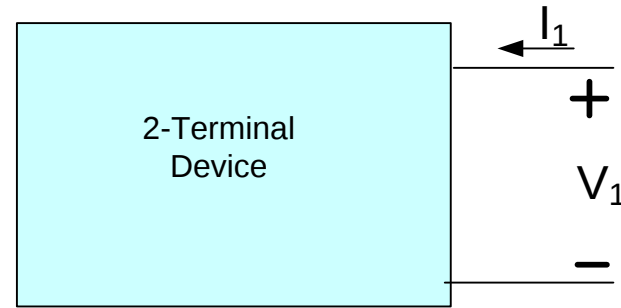


$$\left. \begin{aligned} i_1 &= g_1(v_1, v_2, v_3) \\ i_2 &= g_2(v_1, v_2, v_3) \\ i_3 &= g_3(v_1, v_2, v_3) \end{aligned} \right\}$$

$$\begin{aligned} i_1 &= y_{11}v_1 + y_{12}v_2 + y_{13}v_3 \\ i_2 &= y_{21}v_1 + y_{22}v_2 + y_{23}v_3 \\ i_3 &= y_{31}v_1 + y_{32}v_2 + y_{33}v_3 \end{aligned}$$

$$y_{ij} = \left. \frac{\partial f_i(\mathbf{v}_1, \mathbf{v}_2, \mathbf{v}_3)}{\partial \mathbf{v}_j} \right|_{\bar{\mathbf{v}} = \bar{\mathbf{v}}_Q}$$

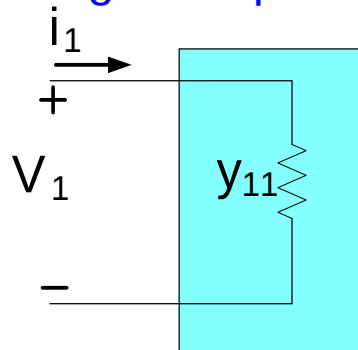
Small-Signal Model



$$\mathbf{i}_1 = \mathbf{y}_{11} \mathbf{v}_1$$

$$y_{11} = \left. \frac{\partial f_1(V_1)}{\partial V_1} \right|_{\bar{V} = \bar{V}_Q} \quad \bar{V} = V_{1Q}$$

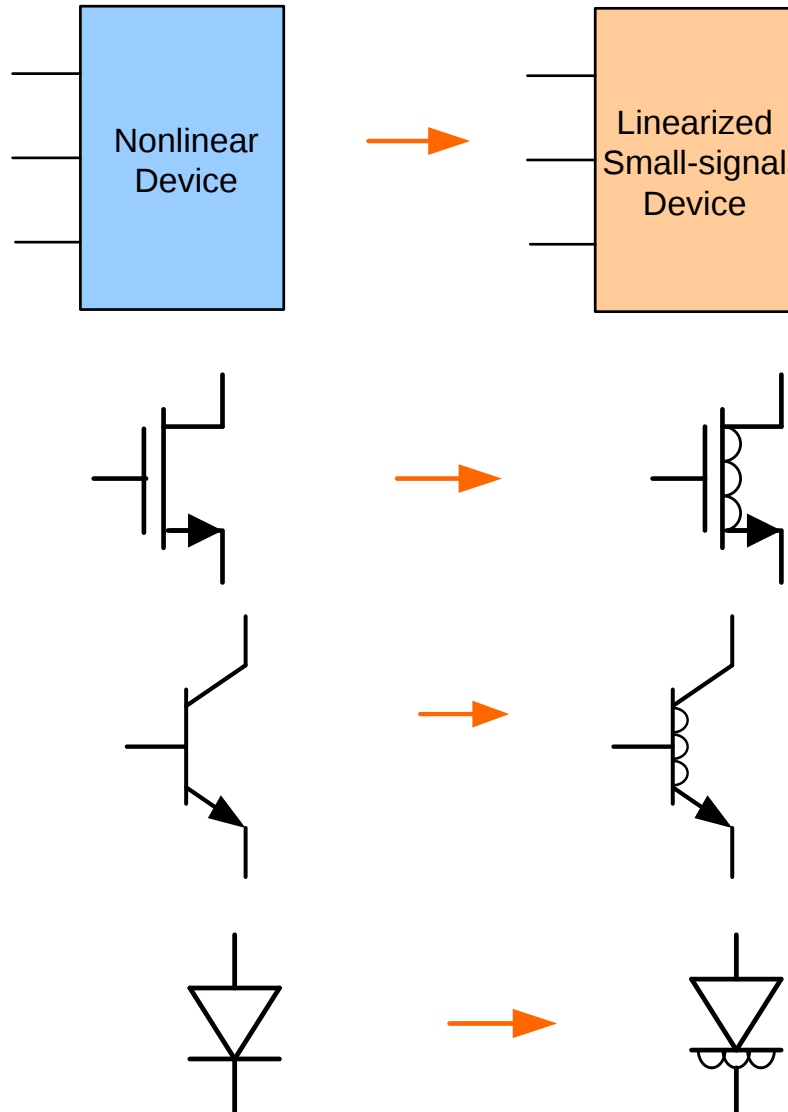
A Small Signal Equivalent Circuit



Small-signal model is a one-port

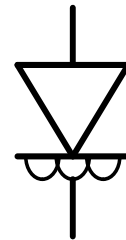
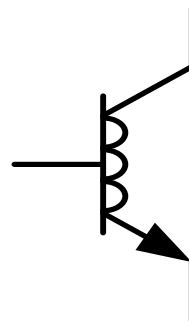
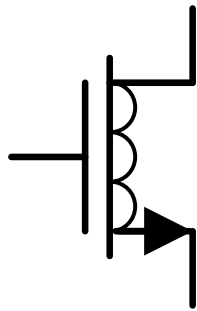
This was actually developed earlier !

Linearized nonlinear devices



How is the small-signal equivalent circuit obtained from the nonlinear circuit?

What is the small-signal equivalent of the MOSFET, BJT, and diode ?



End of Lecture 23