

EE 508

Lecture 29

Integrator Design

- Nonideal Effects in SC Circuits
- Switched Capacitor Amplifiers
- Switched Resistor Filters
- Some other integrator structures
- Metrics for comparing integrators

Nonideal Effects in Switched Capacitor Circuits

- Parasitic Capacitances
- Charge Injection
- Aliasing
- Redundant Switch Removal
- Matching
- • Noise

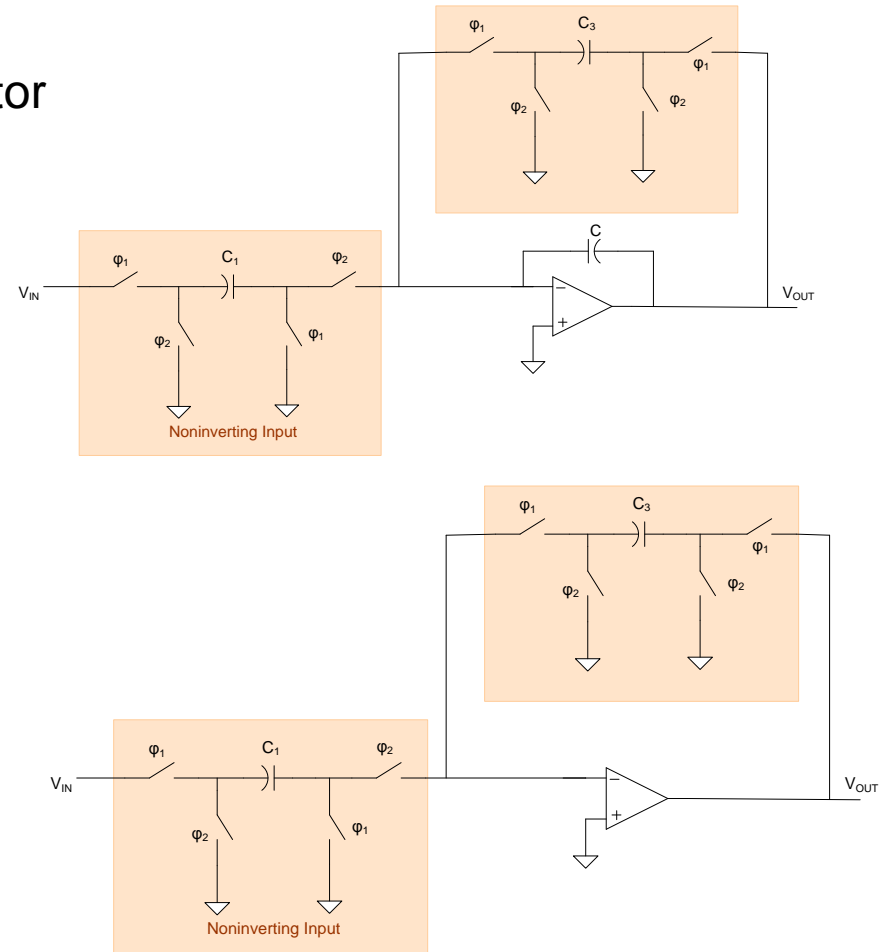
Noise

- Capacitors do not have any noise source
- Switches contribute thermal noise
- Noise due to switches looks like “capacitive” noise $V_{RMS} = \sqrt{\frac{kT}{C}}$

Be careful with calculating noise in SC circuits !

Switched Capacitor Amplifiers

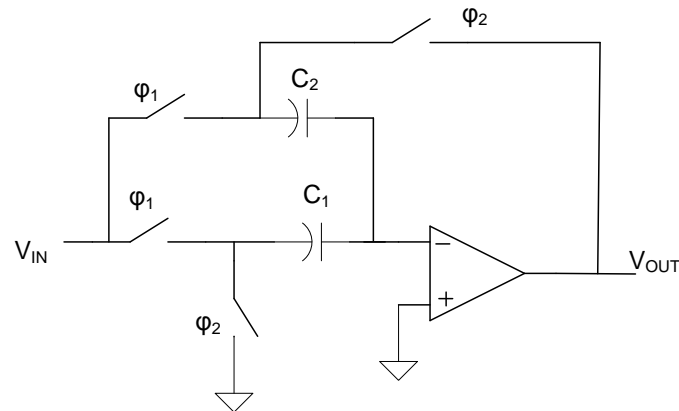
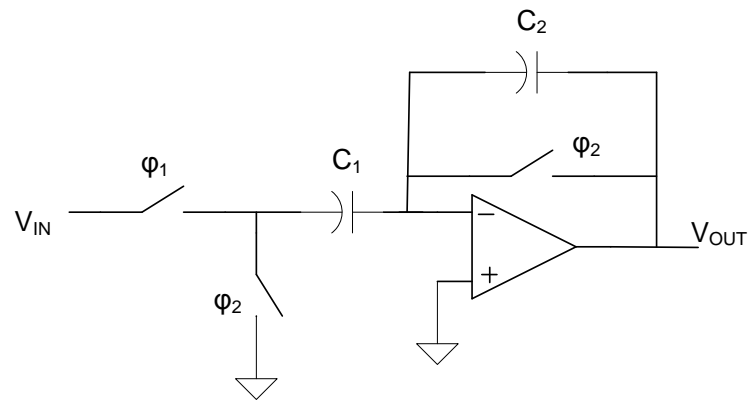
Elimination of the Integration Capacitor



What happens if the integration capacitor is eliminated?

- Serves as a SC amplifier with gain of $A_V = C_1/C_2$
- SC amplifiers and SC summing amplifiers are widely used in filter and non-filter applications

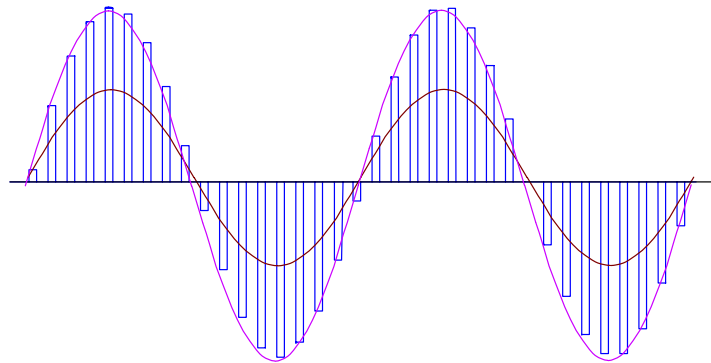
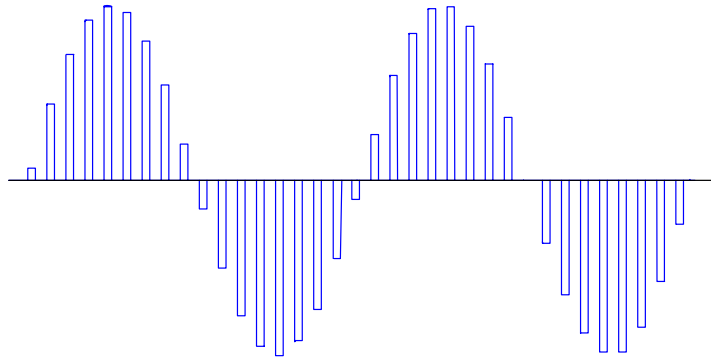
Switched Capacitor Amplifiers



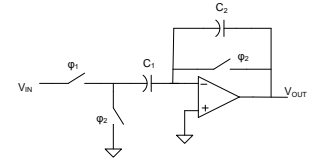
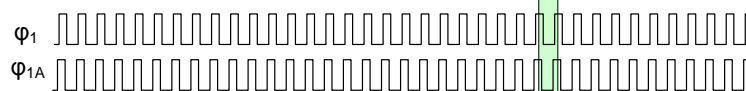
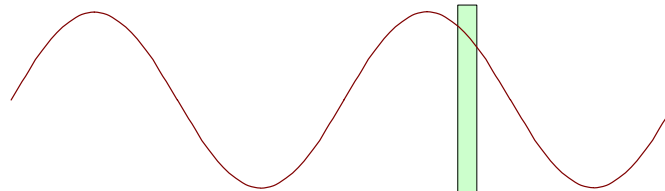
- Summing, Differencing, Inverting, and Noninverting SC Amplifiers Widely Used
- Significant reduction in switches from what we started with by eliminating C in SC integrator
- Must be stray insensitive in most applications
- Outputs valid only during one phase

Switched Capacitor Amplifiers

Output



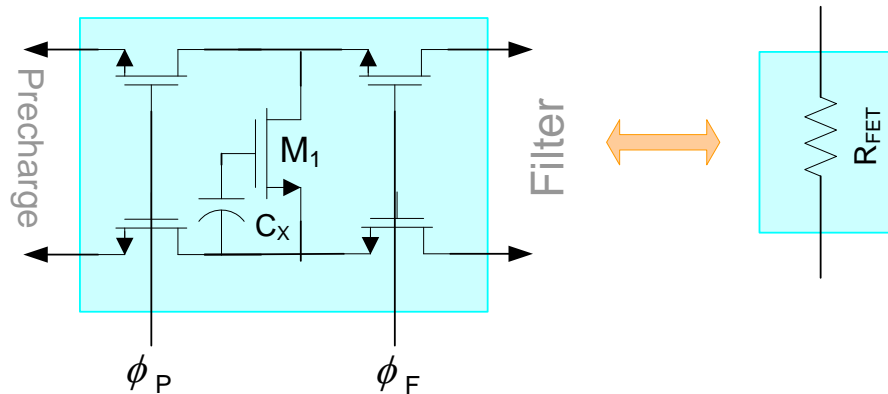
Input



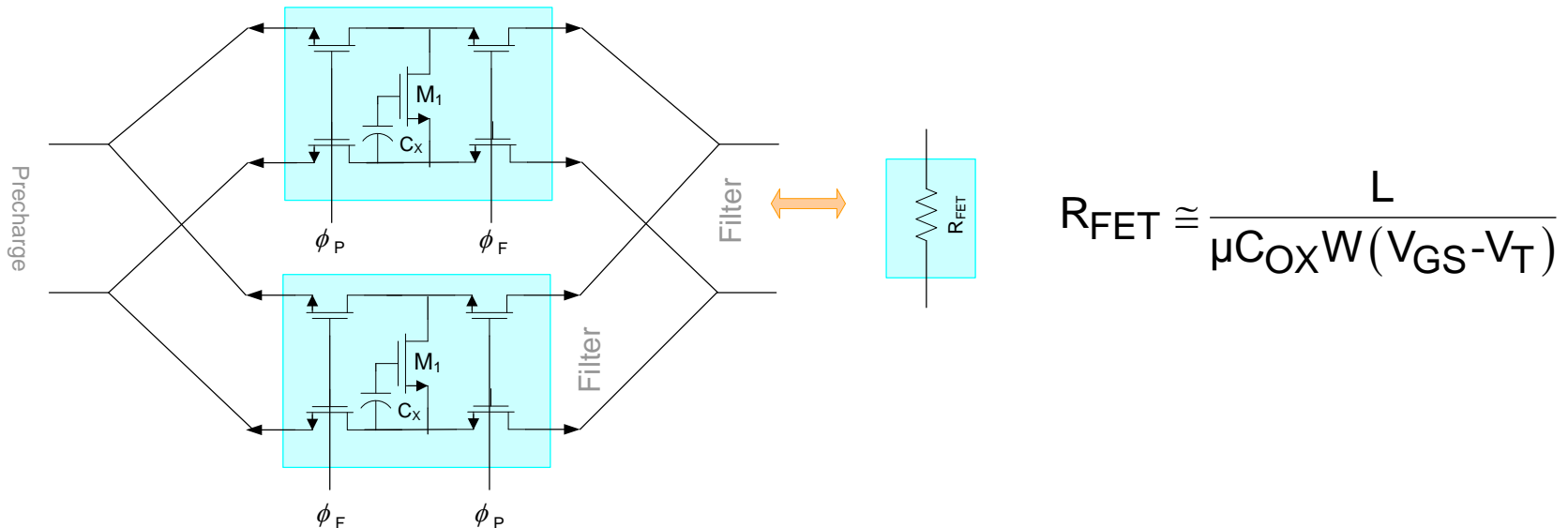
Voltage Mode Integrators

- Active RC (Feedback-based)
 - MOSFET-C (Feedback-based)
 - OTA-C
 - TA-C
 - Switched Capacitor
 - Switched Resistor
 - Other Structures
- Sometimes termed “current mode”
- Will discuss later

Switched-Resistor Voltage Mode Integrators

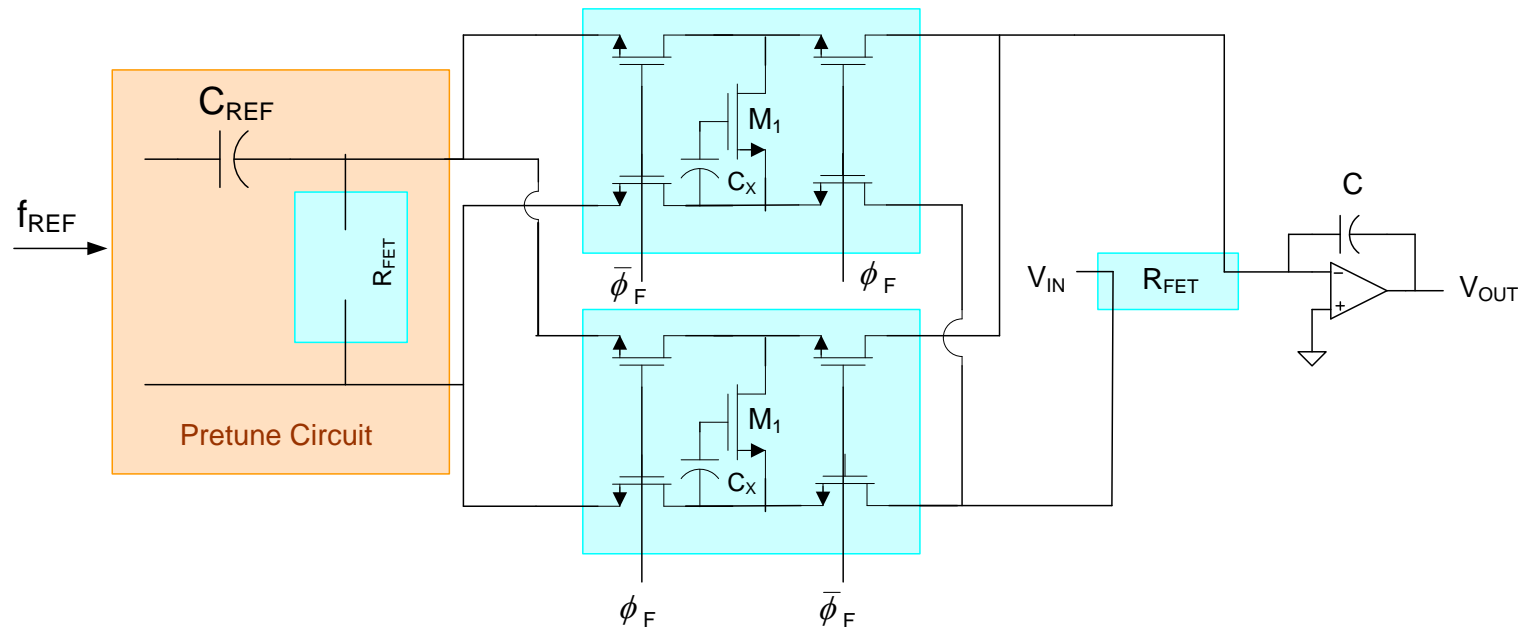


Observe that if a triode-region MOS device is switched between a precharge circuit and a filter circuit (or integrator) and V_{GS} is held constant, It will behave as a resistor while in the filter circuit



Observe that if two such circuits are switched between a precharge circuit and a filter circuit (or integrator) and V_{GS} is held constant, It will behave as a resistor in the filter circuit at all times

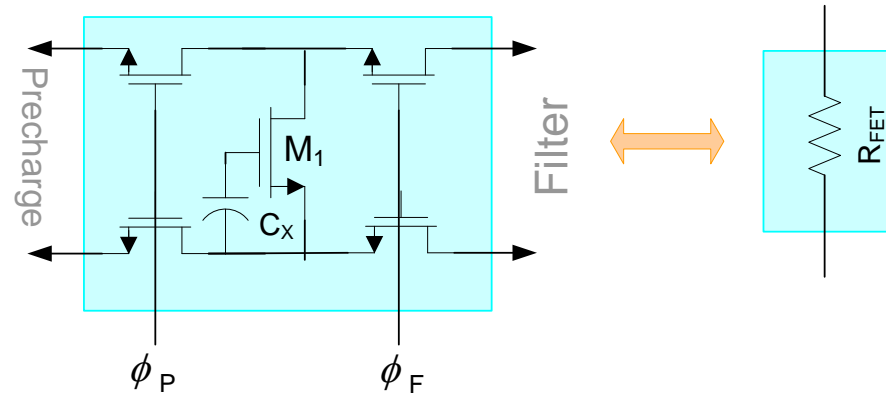
Switched-Resistor Voltage Mode Integrators



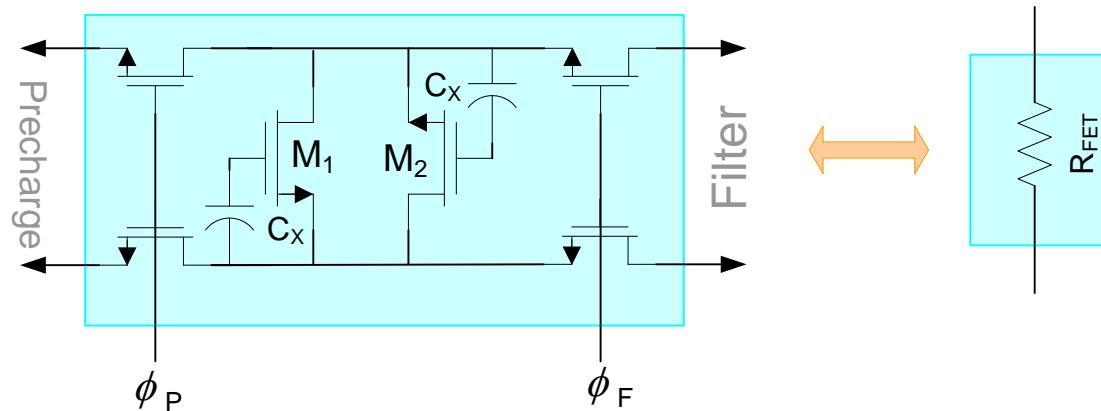
Switched-resistor integrator

- Clock frequency need only be fast enough to prevent droop on C_X
- Minor overlap or non-overlap of clock plays minimal role in integrator performance
- Switched-resistors can be used for integrator resistor or to replace all resistors in any filter
- Pretune circuit can accurately establish $R_{FET} C_{REF}$ product proportional to f_{REF}
- $R_{FET} C$ product is given by accurately controlled $R_{FET} C = R_{FET} C \frac{C_{REF}}{C_{REF}} = [R_{FET} C_{REF}] \cdot \left[\frac{C}{C_{REF}} \right]$ and is thus

Switched-Resistor Voltage Mode Integrators

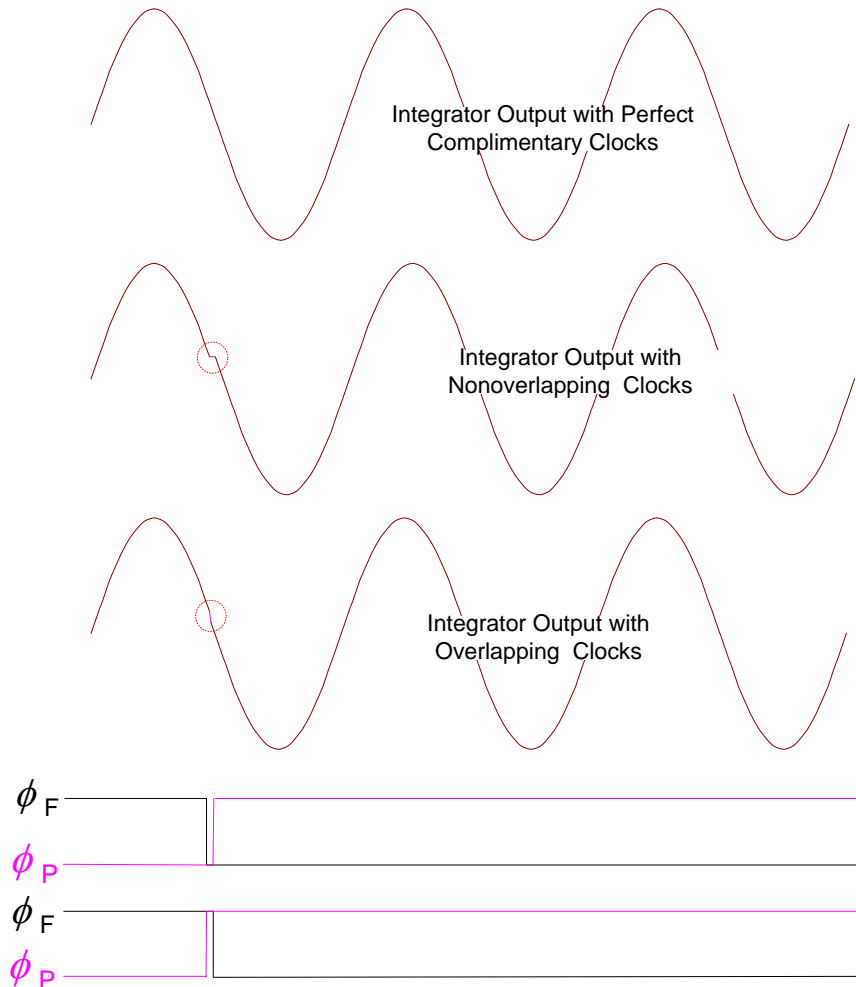


There are some modest nonlinearities in this MOSFET when operating in the triode region



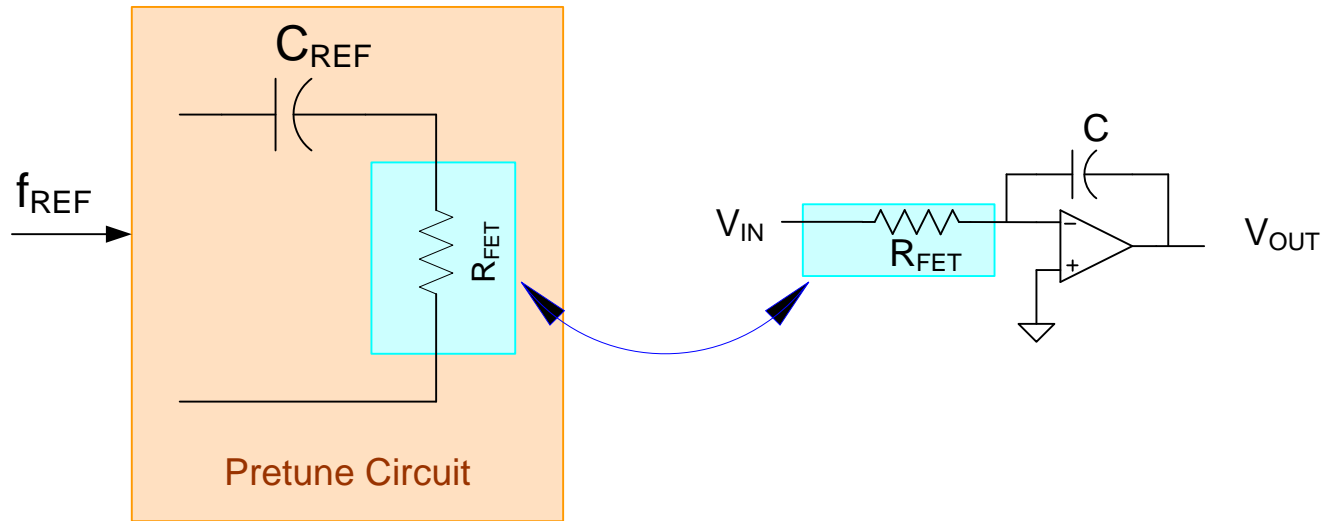
- Significant improvement in linearity by cross-coupling a pair of triode region resistors
- Perfectly cancels nonlinearities if square law model is valid for M_1 and M_2
- Only modest additional complexity in the Precharge circuit

Switched-Resistor Voltage Mode Integrators



- Aberrations are very small, occur very infrequently, and are further filtered
- Play almost no role on performance of integrator or filter

Switched-Resistor Voltage Mode Integrators



Switched-resistor integrator

- Accurate CR_{FET} products is possible
- Area reduced compared to Active RC structure because R_{FET} small
- Single pretune circuit can be used to “calibrate” large number of resistors
- Clock frequency not fast and not critical (but accuracy of f_{REF} is important)
- Since resistors are memoryless elements, no transients associated with switching
- Since filter is a feedback structure, speed limited by BW of op amp

Voltage Mode Integrators

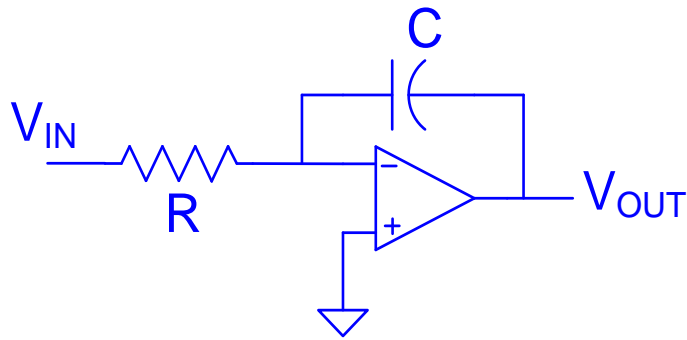
- Active RC (Feedback-based)
 - MOSFET-C (Feedback-based)
 - OTA-C
 - TA-C
 - Switched Capacitor
 - Switched Resistor
 - Other Structures
- Sometimes termed “current mode”
- Will discuss later

Have introduced a basic voltage-mode integrators in each of these approaches

All of these structures have applications where they are useful

Performance of all are limited by variability or Op Amp BW

Variants of basic inverting integrator have been considered



Basic Miller Integrator

- Active RC
- MOSFET-C
- OTA-C
- g_m -C
- Switched-Capacitor
- Switched-Resistor

Performance of all is limited by GB of Operational Amplifiers

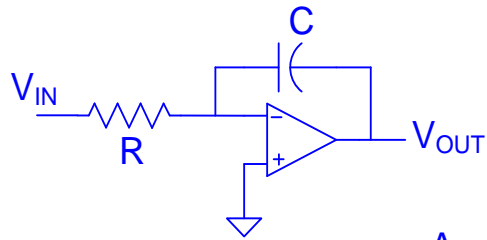
How can integrator performance be improved?

- Better op amps
- Better Integrator Architectures

How can the performance of integrator structures be compared?

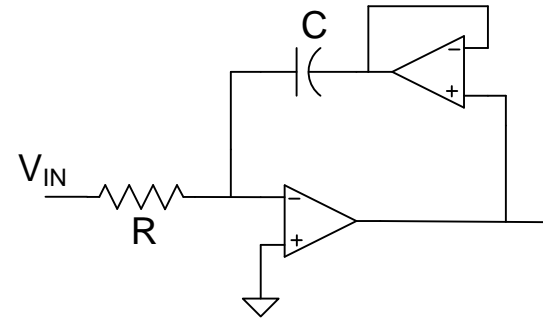
Need metric for comparing integrator performance

Are there other integrators in the basic classes that have been considered?



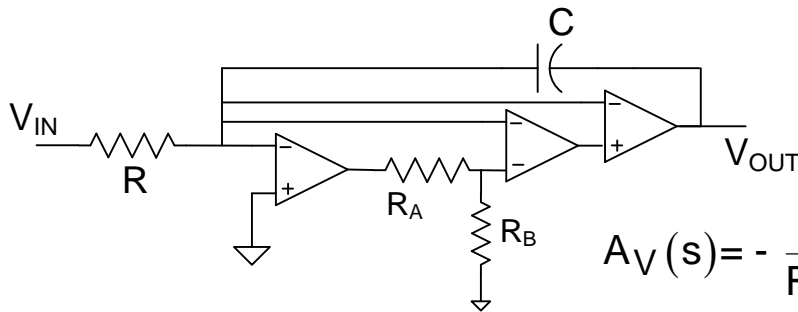
Miller Inverting

$$A_V(s) = -\frac{1}{RCs}$$



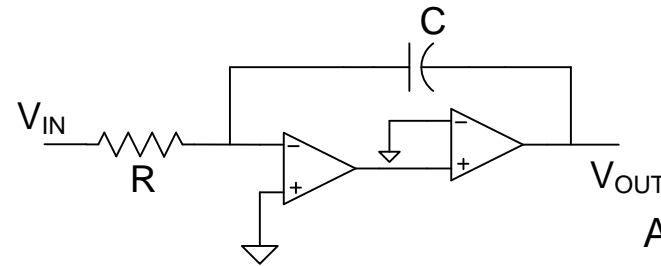
High-Q Inverting

$$A_V(s) = -\frac{1}{RCs}$$



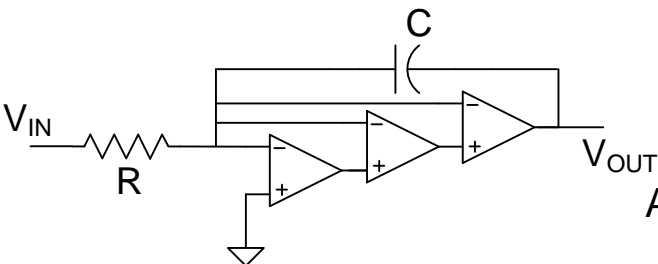
Zero Second Derivative Inverting

$$A_V(s) = -\frac{1}{RCs}$$



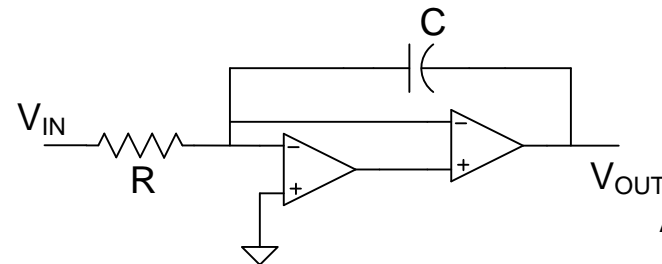
Cascaded Inverting

$$A_V(s) = -\frac{1}{RCs}$$



Zero Second Derivative Inverting

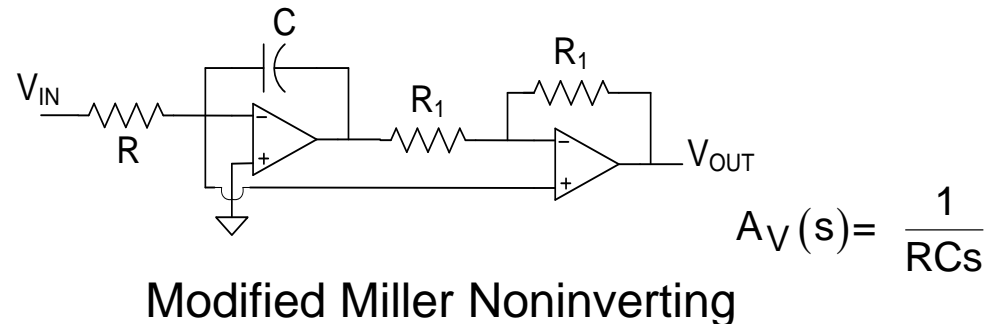
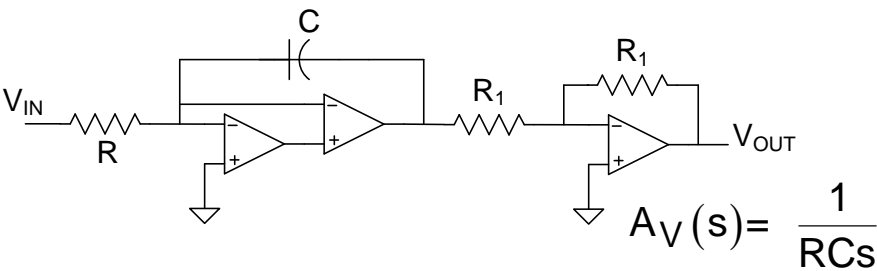
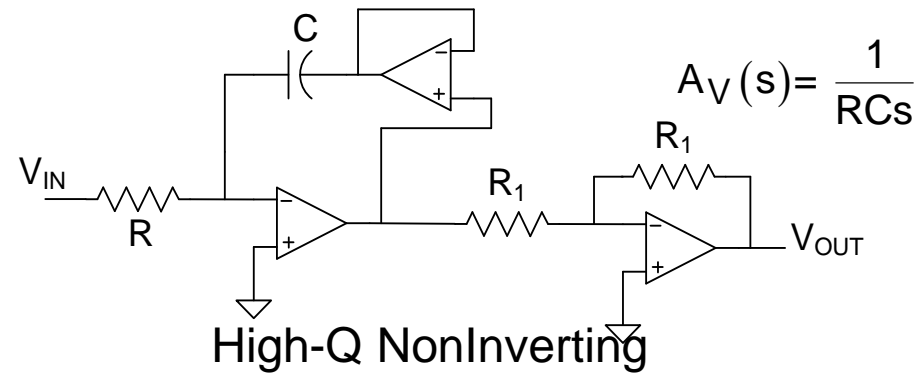
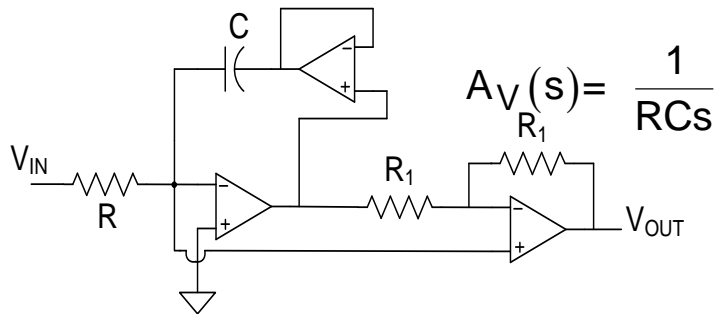
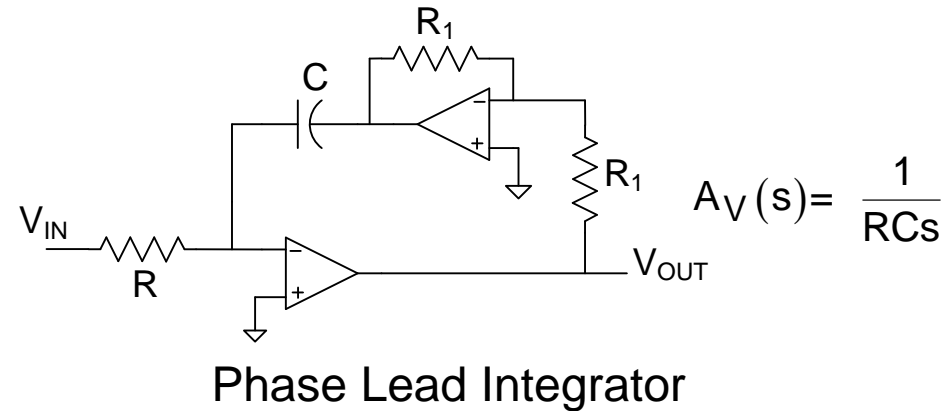
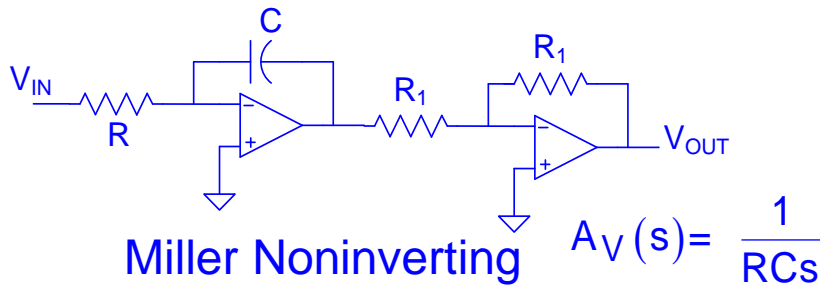
$$A_V(s) = -\frac{1}{RCs}$$



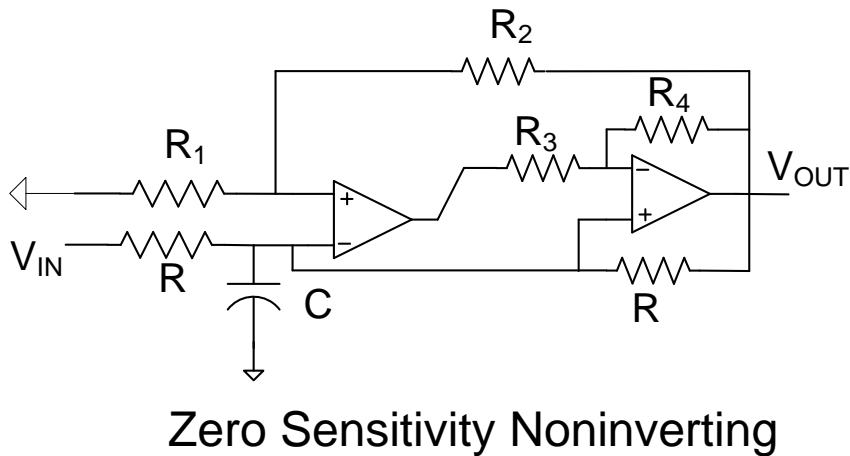
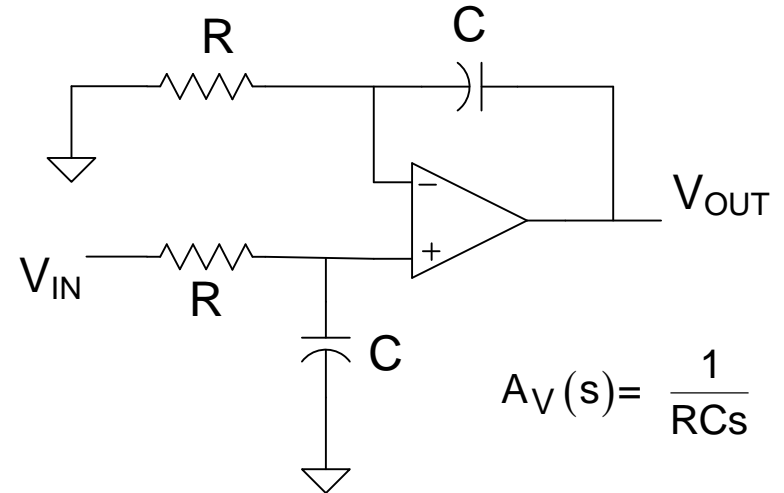
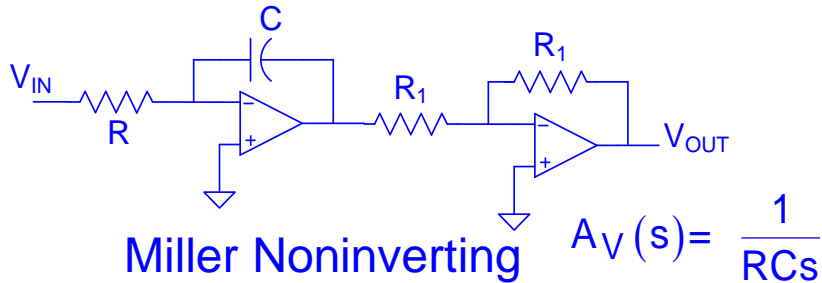
Zero Sensitivity Inverting

$$A_V(s) = -\frac{1}{RCs}$$

Are there other integrators in the basic classes that have been considered?



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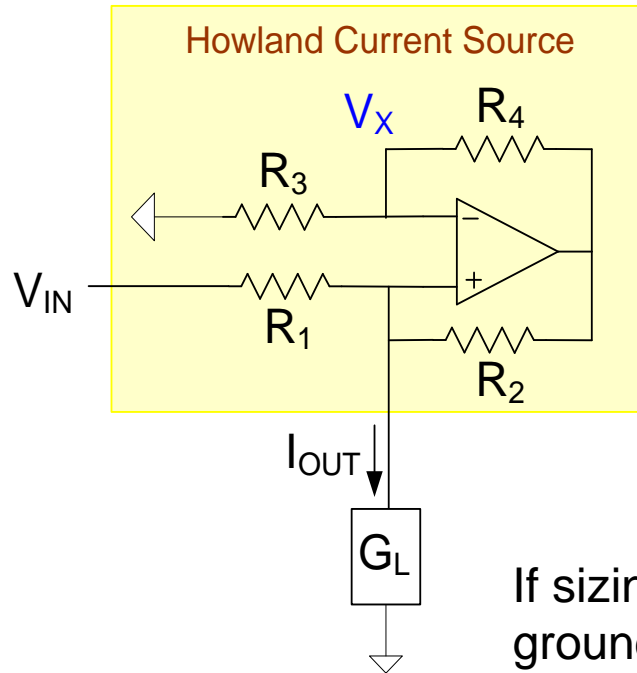
$$A_V(s) = \frac{2}{RCs}$$

If $R_1=R_2$ and $R_3=R_4$

(note this has a grounded integrating capacitor!)

De Boo Integrator

Consider the Howland Current Source



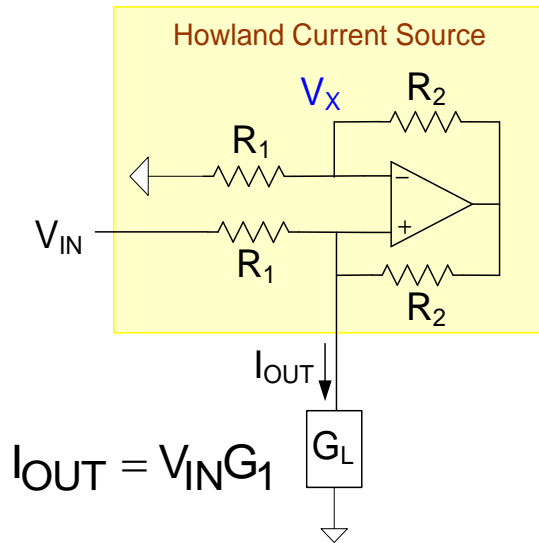
$$I_{OUT} = V_{IN}G_1 + \left[V_X \left(\frac{G_2G_3}{G_4} - G_1 \right) \right]$$

If resistors sized so that $G_1 = \frac{G_2G_3}{G_4}$

$$I_{OUT} = V_{IN}G_1$$

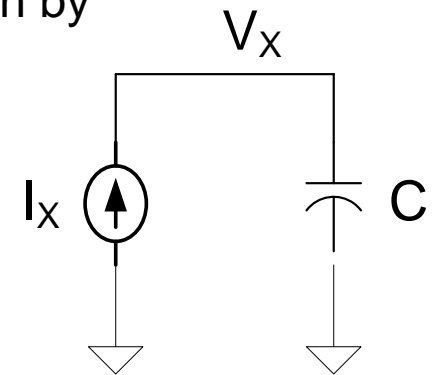
If sizing constraints are satisfied, behaves as a grounded constant-current source

DeBoo Integrator



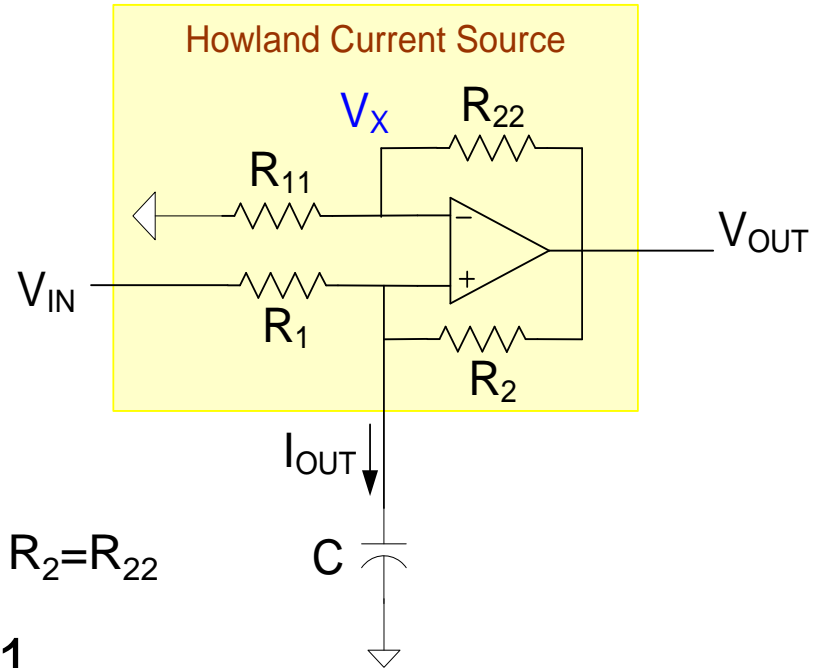
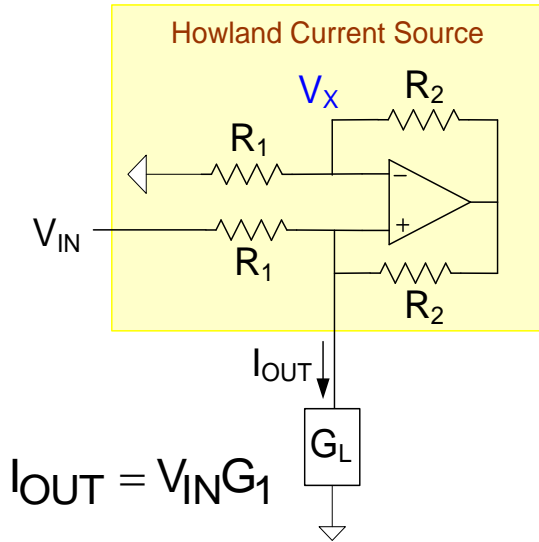
Observe that if a current source drives a grounded capacitor, then the nodal voltage on the capacitor is given by

$$V_X = I_X \frac{1}{sC}$$



Thus, if we could make I_X proportional to V_{IN} , the voltage on the capacitor would be a weighted Integral of V_{IN}

De Boo Integrator



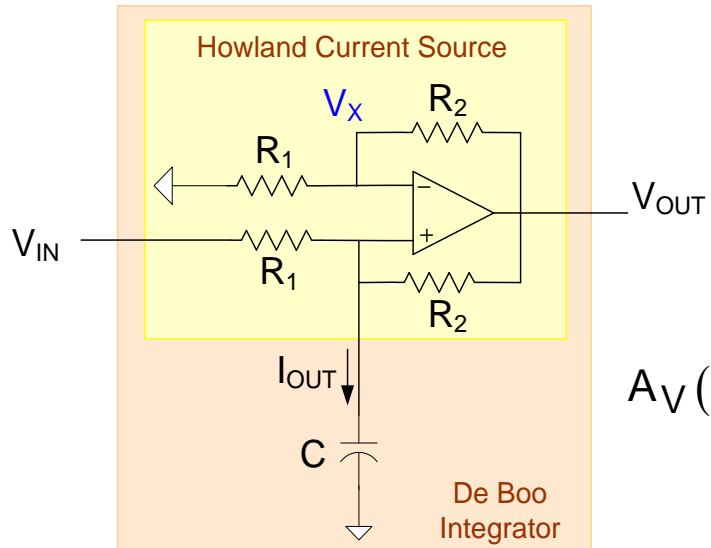
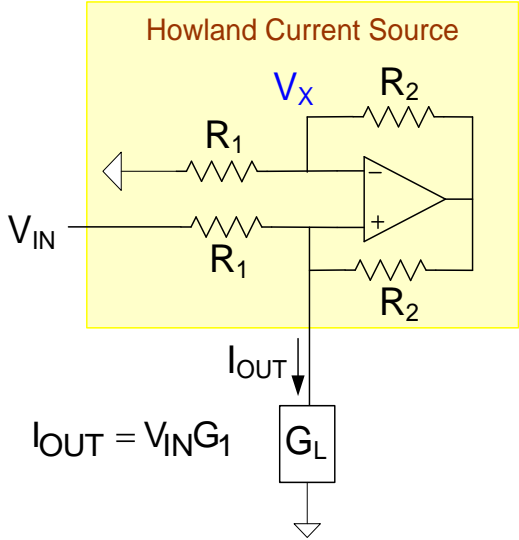
If $R_1=R_{11}$ and $R_2=R_{22}$

$$V_X = \frac{V_{IN}}{R_1} \frac{1}{sC}$$

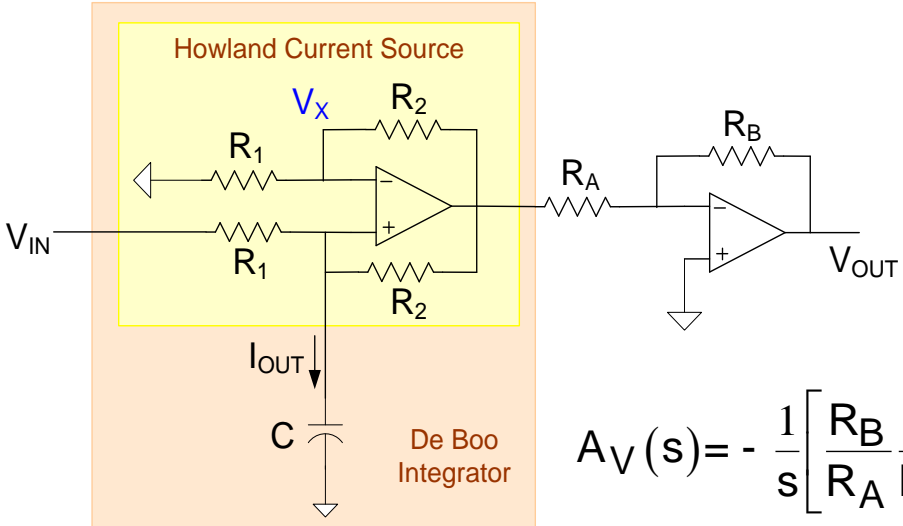
$$V_{OUT} = V_X \left(1 + \frac{R_{22}}{R_{11}} \right) = \frac{V_{IN}}{R_1} \frac{1}{sC} \left(1 + \frac{R_{22}}{R_{11}} \right)$$

$$A_V(s) = \frac{1}{s} \left[\frac{1}{R_1 C} \left(1 + \frac{R_{22}}{R_{11}} \right) \right]$$

De Boo Integrator



$$A_V(s) = \frac{1}{s} \left[\frac{1}{R_1 C} \left(1 + \frac{R_{22}}{R_{11}} \right) \right]$$



$$A_V(s) = - \frac{1}{s} \left[\frac{R_B}{R_A} \frac{1}{R_1 C} \left(1 + \frac{R_{22}}{R_{11}} \right) \right]$$

Many different integrator architectures that ideally provide the same gain

Similar observations can be made for other classes of integrators

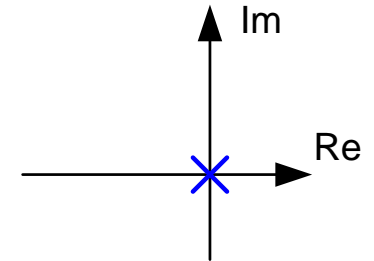
How can the performance of an integrator be characterized and how can integrators be compared?

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Consider Ideal Integrator Gain Function

$$A_V(s) = \frac{I_0}{s} \quad A_V(j\omega) = \frac{I_0}{j\omega}$$

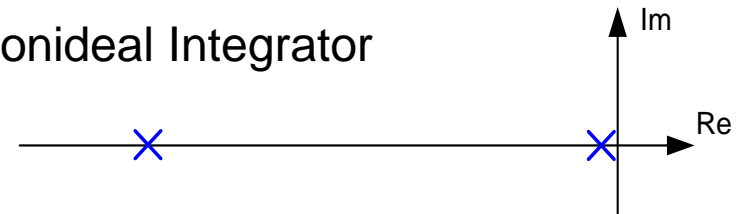
Ideal Integrator



Consider a nonideal integrator Gain Function

$$A_V(s) = \frac{\alpha I_{01}}{s + \alpha} A_{OO}(s)$$

Nonideal Integrator



Key characteristics of an ideal integrator:

- Magnitude of the gain at $I_0=1$
- Phase of integrator always 90°
- Gain decreases with $1/\omega$

Are any of these properties more critical than others?

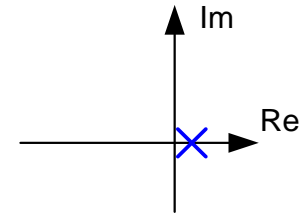
In many applications:

Key property of ideal integrator is a phase shift of 90° at frequencies around I_0 !

How can the performance of an integrator be characterized and how can integrators be compared?

Is stability of an integrator of concern?

Ideal Integrator



- Ideal integrator is not stable
- Integrator function is inherently ill-conditioned
- Integrator is almost never used open-loop
- Stability of integrator not of concern, stability of filter using integrator is of concern
- Some integrators may cause unstable filters, others may result in stable filters
- Instability in filter because desired poles move in RHP is of little concern since the filter performance would be unacceptable long before the stability became an issue
- Instability in filter due to parasitic poles is of concern but not a problem in most circuits

End of Lecture 29